

Controller Architectures: Tradeoffs between Performance and Complexity

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NREL Autonomous Energy Grids Workshop

Inter-area oscillations in power systems

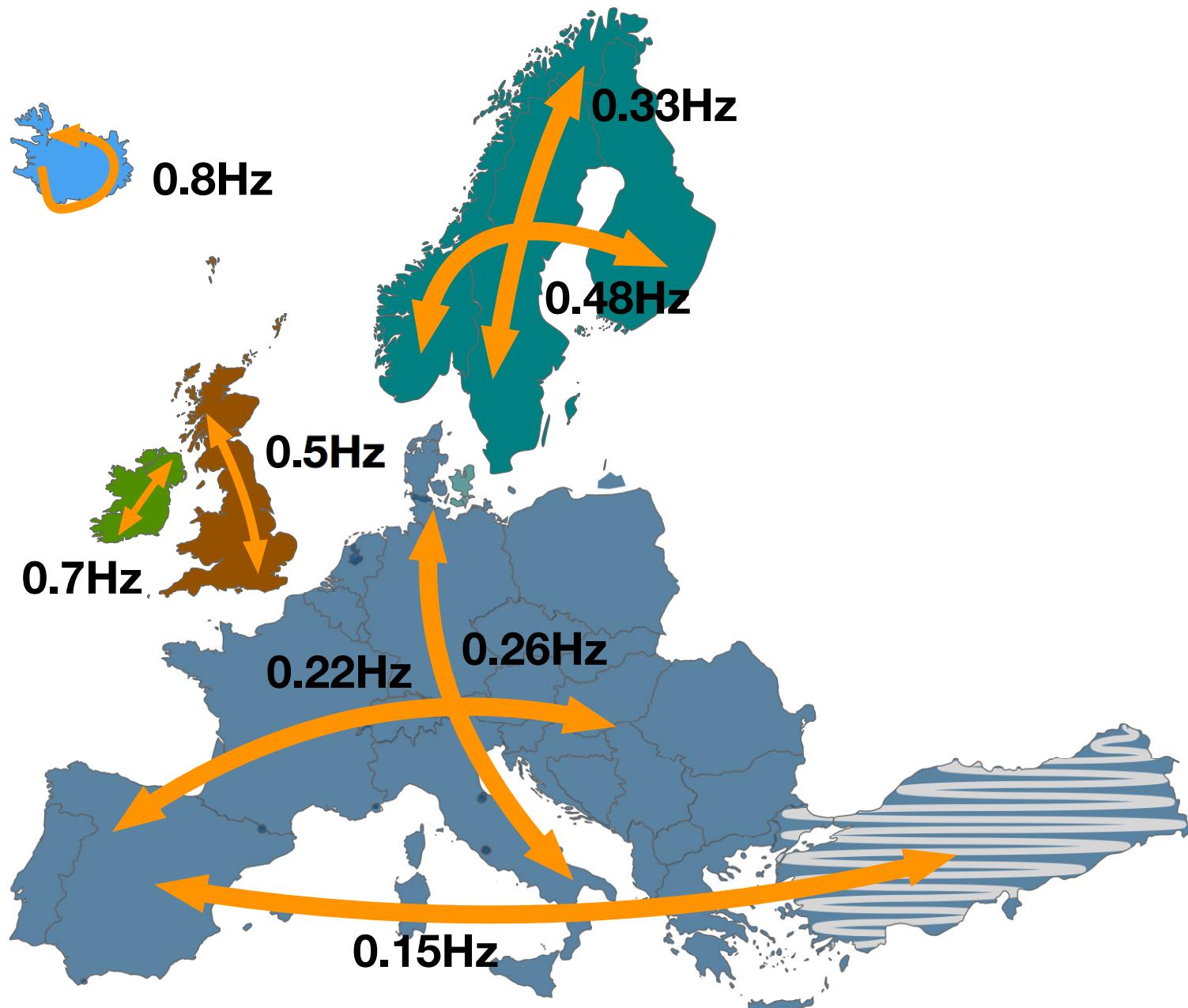
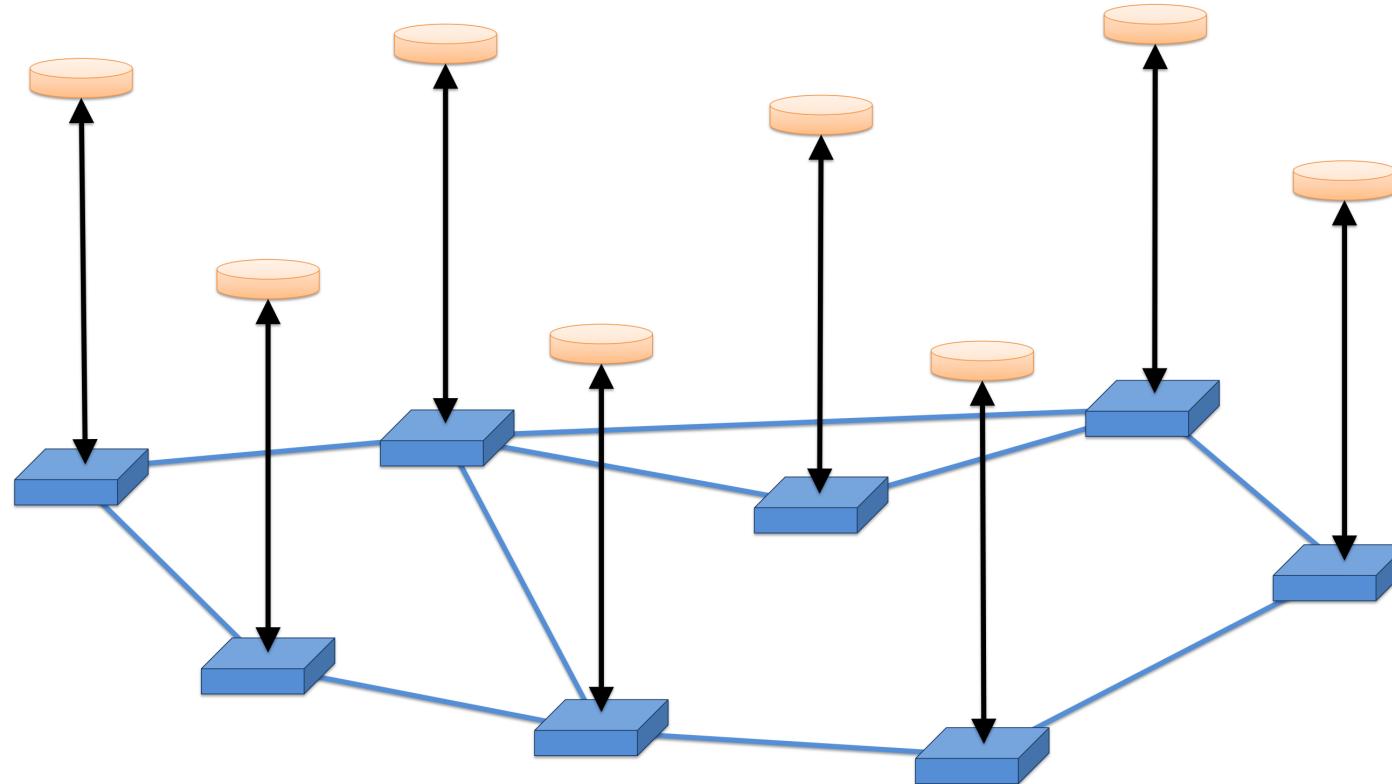


Image credit: Florian Dörfler

Conventional control of generators

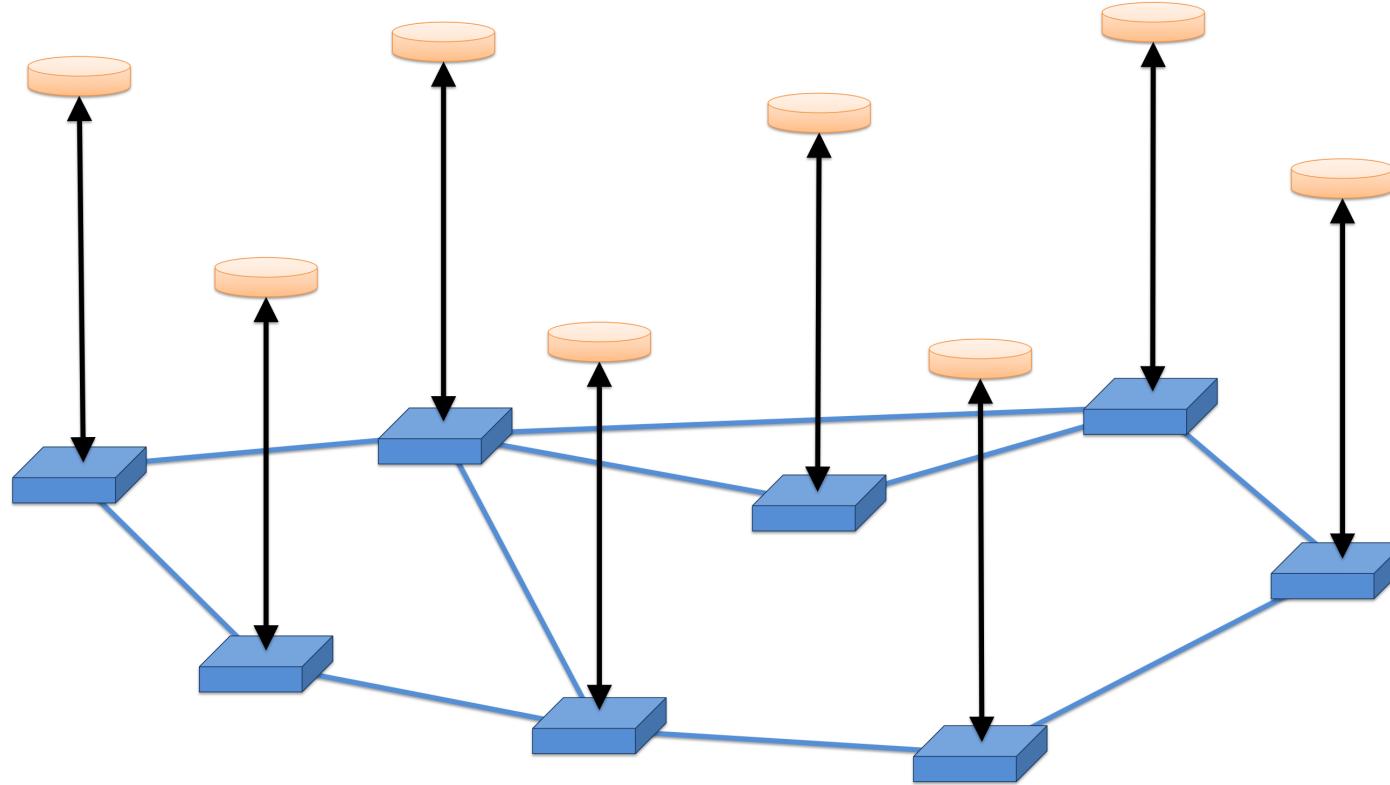
fully decentralized controller



network of generators

Conventional control of generators

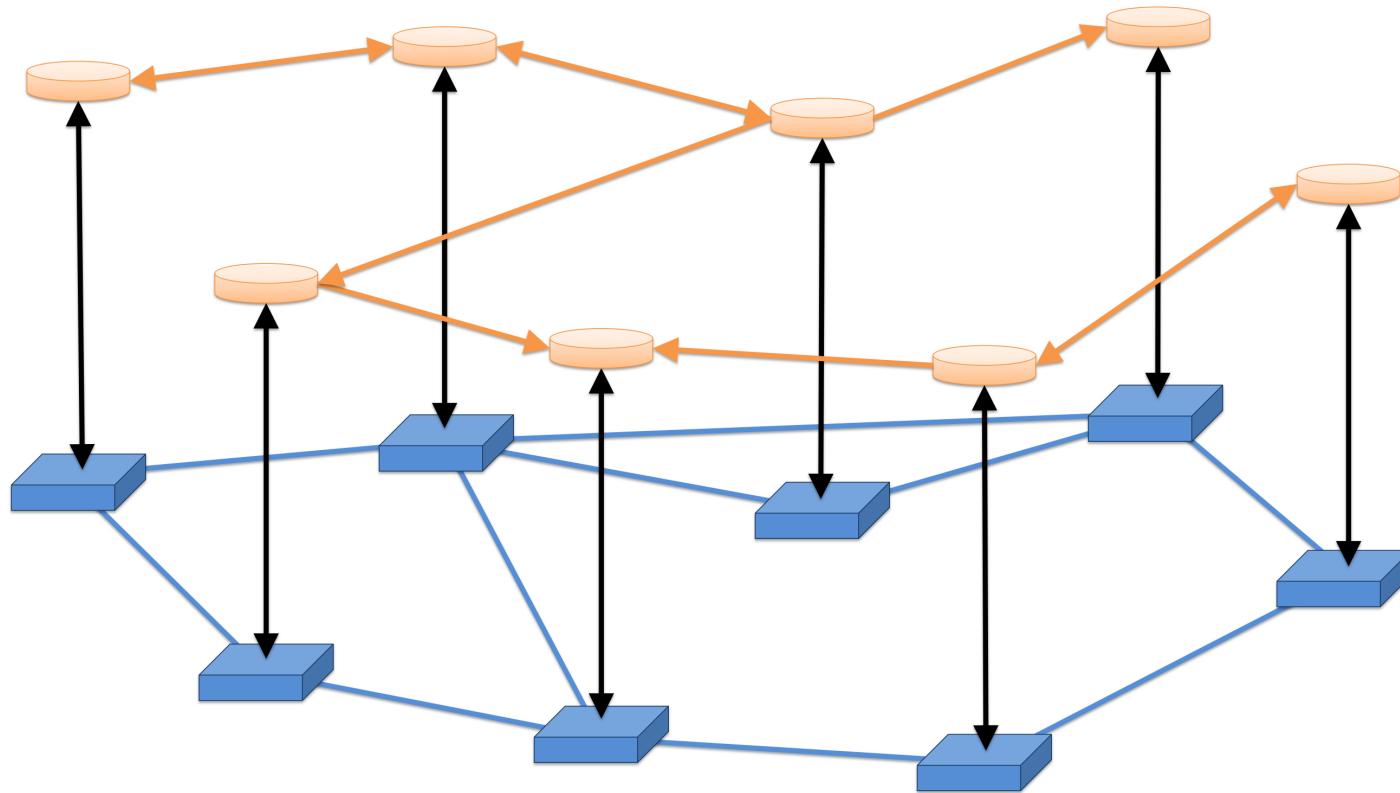
fully decentralized controller



network of generators

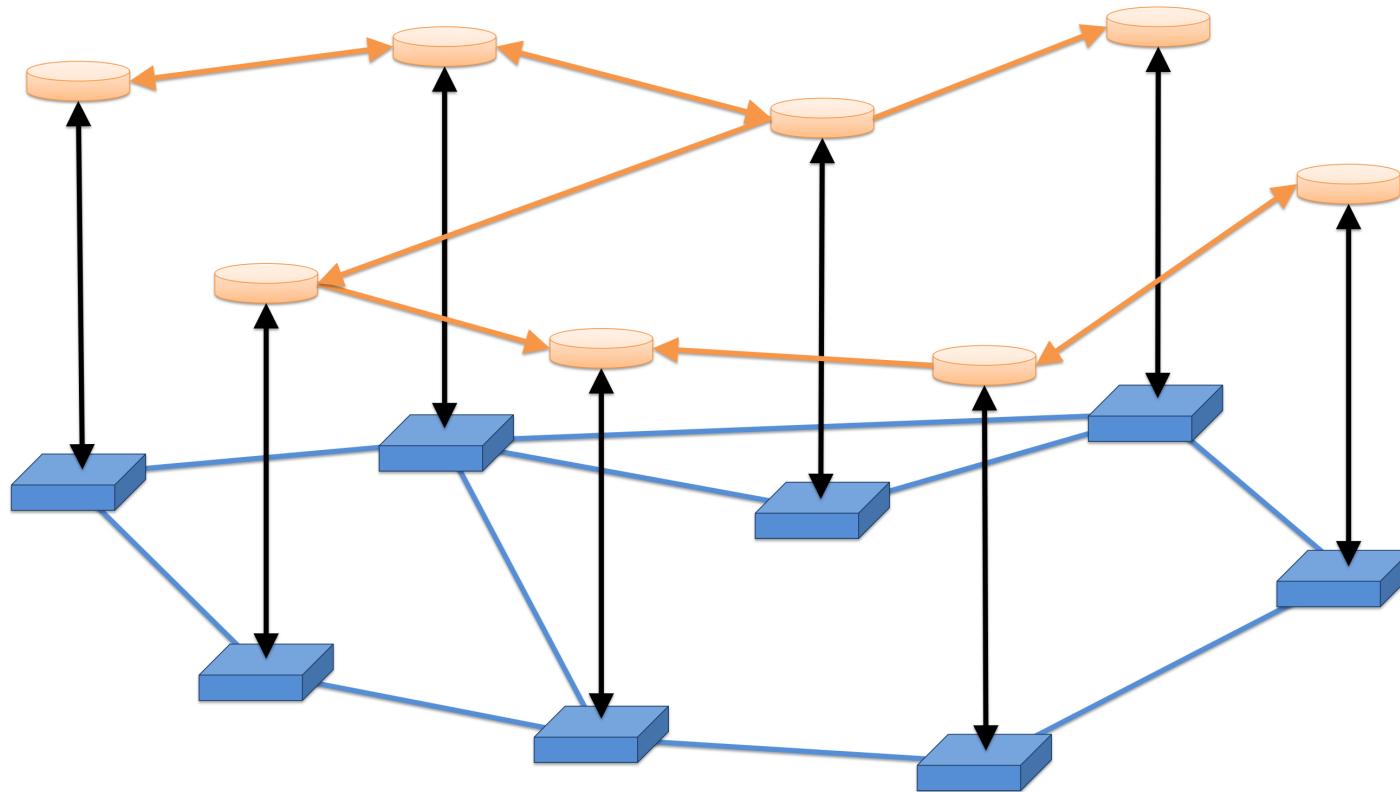
- CONVENTIONAL CONTROL
 - * local oscillations ✓
 - * inter-area oscillations ✗

Possible alternative structured dynamic controller



distributed plant and its interaction links

Possible alternative structured dynamic controller



distributed plant and its interaction links

CHALLENGE

design of controller architectures
performance vs complexity

Complexity via Regularization

minimize

$$J(K)$$

+

$$\gamma g(K)$$



**closed-loop
performance**

**controller
complexity**

$\gamma > 0$ – performance vs complexity tradeoff

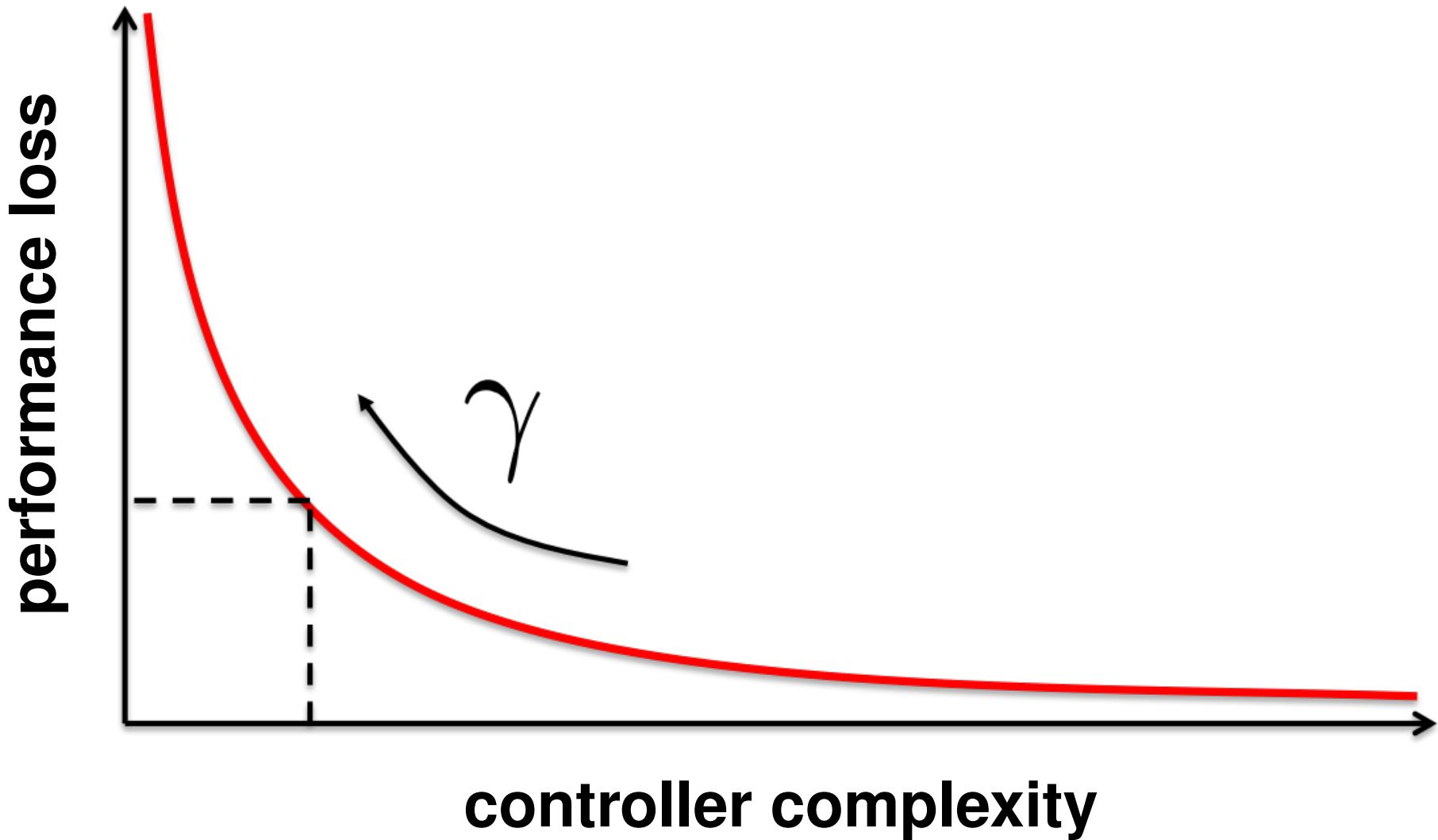
Fardad, Lin, Jovanović, ACC '11

Lin, Fardad, Jovanović, IEEE TAC '13

Matni & Chandrasekaran, IEEE TAC '16

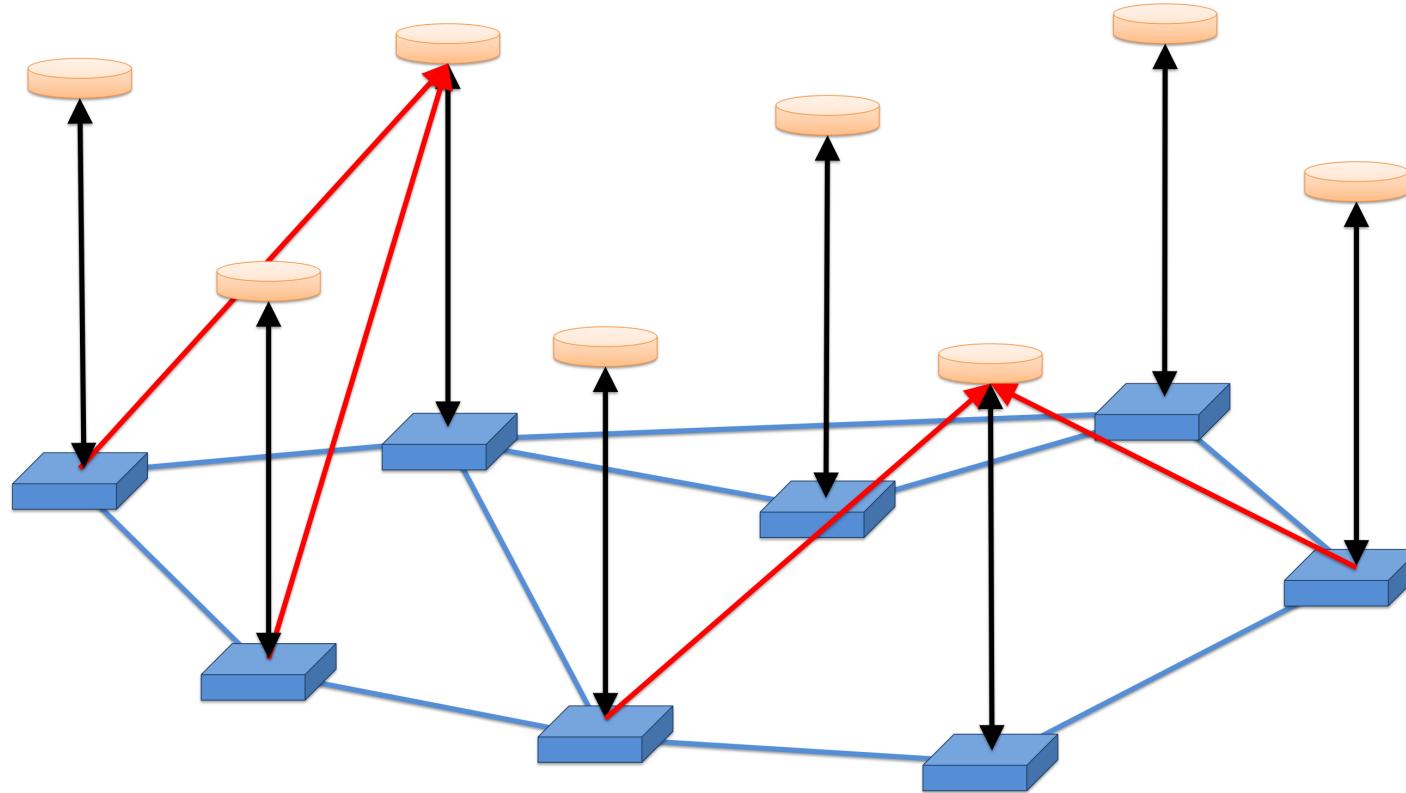
- TRADE-OFF CURVE

- ★ **performance vs complexity**



This talk

structured memoryless controller



distributed plant and its interaction links

OBJECTIVE

identification of a **signal exchange network**
performance vs sparsity

CONTROL PROBLEM

Lyapunov equation

discrete-time dynamics: $x_{t+1} = A x_t + B d_t$

white-in-time forcing: $\mathbf{E}(d_t d_\tau^T) = W \delta_{t-\tau}$

- LYAPUNOV EQUATION

$$\begin{aligned}
 \textcolor{red}{X}_{t+1} &:= \mathbf{E}(x_{t+1} x_{t+1}^T) \\
 &= \mathbf{E}((A x_t + B d_t)(x_t^T A^T + d_t^T B^T)) \\
 &= A \mathbf{E}(x_t x_t^T) A^T + B \mathbf{E}(d_t d_t^T) B^T \\
 &= A \textcolor{red}{X}_t A^T + B \textcolor{blue}{W} B^T
 \end{aligned}$$

- ★ continuous-time version

$$\frac{d \textcolor{red}{X}_t}{d t} = A \textcolor{red}{X}_t + \textcolor{red}{X}_t A^T + B \textcolor{blue}{W} B^T$$

Minimum variance state-feedback problem

dynamics: $\dot{x} = Ax + B_1 d + B_2 u$

objective function: $J = \lim_{t \rightarrow \infty} \mathbf{E} (x^T(t) Q x(t) + u^T(t) R u(t))$

memoryless controller: $u = -\textcolor{red}{F} x$

Minimum variance state-feedback problem

dynamics: $\dot{x} = Ax + B_1 d + B_2 u$

objective function: $J = \lim_{t \rightarrow \infty} \mathbf{E} (x^T(t) Q x(t) + u^T(t) R u(t))$

memoryless controller: $u = -\textcolor{red}{F} x$

- CLOSED-LOOP VARIANCE

J – **non-convex** function of F

No structural constraints

- SDP CHARACTERIZATION

$$\underset{X, F}{\text{minimize}} \quad \text{trace} \left((Q + F^T R F) X \right)$$

$$\text{subject to } (A - B_2 F) X + X (A - B_2 F)^T + B_1 B_1^T = 0$$

$$X \succ 0$$

No structural constraints

- SDP CHARACTERIZATION

$$\underset{X, F}{\text{minimize}} \quad \text{trace} \left((Q + F^T R F) X \right)$$

$$\text{subject to } (A - B_2 F) X + X (A - B_2 F)^T + B_1 B_1^T = 0$$

$$X \succ 0$$

* change of variables: $FX = Y$

$$\underset{X, Y}{\text{minimize}} \quad \text{trace}(Q X) + \text{trace}(R Y X^{-1} Y^T)$$

$$\text{subject to } (A X - B_2 Y) + (A X - B_2 Y)^T + B_1 B_1^T = 0$$

$$X \succ 0$$

Schur complement \Rightarrow SDP characterization

- RICCATI-BASED-CHARACTERIZATION

globally optimal controller

$$A^T \mathcal{P} + \mathcal{P} A - \mathcal{P} B_2 R^{-1} B_2^T \mathcal{P} + Q = 0$$

$$F_c = R^{-1} B_2^T \mathcal{P}$$

- STRUCTURAL CONSTRAINTS $F \in \mathcal{S}$

centralized

$$\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}$$

fully-decentralized

$$\begin{bmatrix} * & & & \\ & * & & \\ & & * & \\ & & & * \end{bmatrix}$$

localized

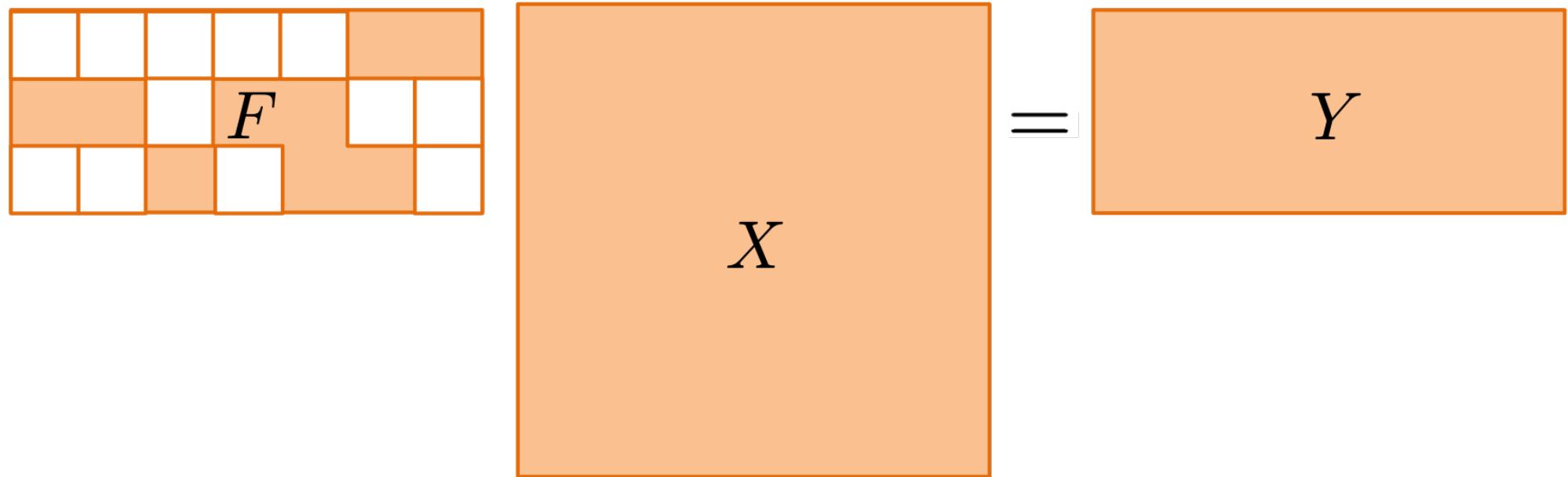
$$\begin{bmatrix} * & * & & \\ * & * & * & \\ & * & * & * \\ & & * & * \end{bmatrix}$$

GRAND CHALLENGE

convex characterization in the face of structural constraints

difficult to establish **relation between**

$\left\{ \begin{array}{l} \text{structural constraints} \\ \text{on } F \end{array} \right\}$ and $\left\{ \begin{array}{l} \text{structural constraints} \\ \text{on } X \text{ and } Y \end{array} \right\}$



Classes of convex problems

- PARTIALLY-NESTED SYSTEMS

Ho & Chu, IEEE TAC '72

Voulgaris, ACC '00; ACC '01

- CONE- AND FUNNEL-CAUSAL SYSTEMS

Voulgaris, Bianchini, Bamieh, SCL '03

Bamieh & Voulgaris, SCL '05

Fardad & Jovanović, Automatica '11

- QUADRATICALLY-INVARIANT SYSTEMS

Rotkowitz & Lall, IEEE TAC '06

- POSET-CAUSAL SYSTEMS

Shah & Parrilo, IEEE TAC '13

- POSITIVE SYSTEMS

Tanaka & Langbort, IEEE TAC '11

Colaneri, Middleton, Chen, Caporale, Blanchini, Automatica '14

Rantzer, EJC '15; IEEE TAC '16

An example



$$u(t) = - [F_p \quad F_v] \begin{bmatrix} p(t) \\ v(t) \end{bmatrix}$$

- **OBJECTIVE**

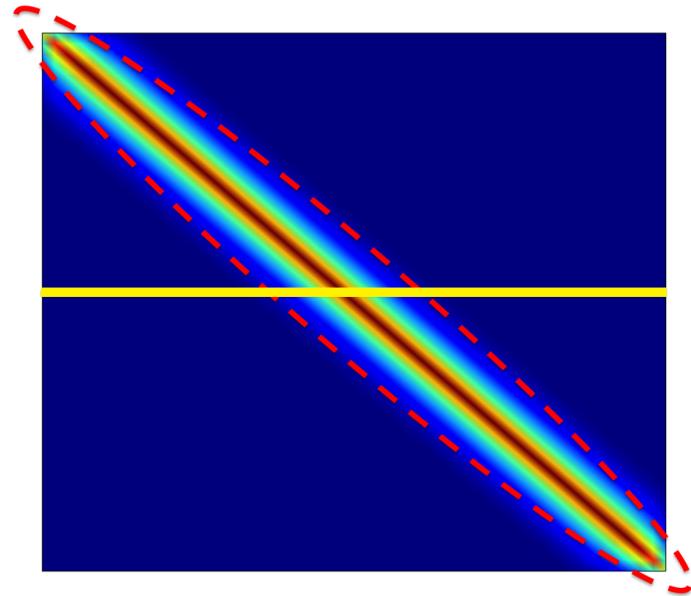
- ★ minimize steady-state variance of p, v, u

optimal controller – Linear Quadratic Regulator

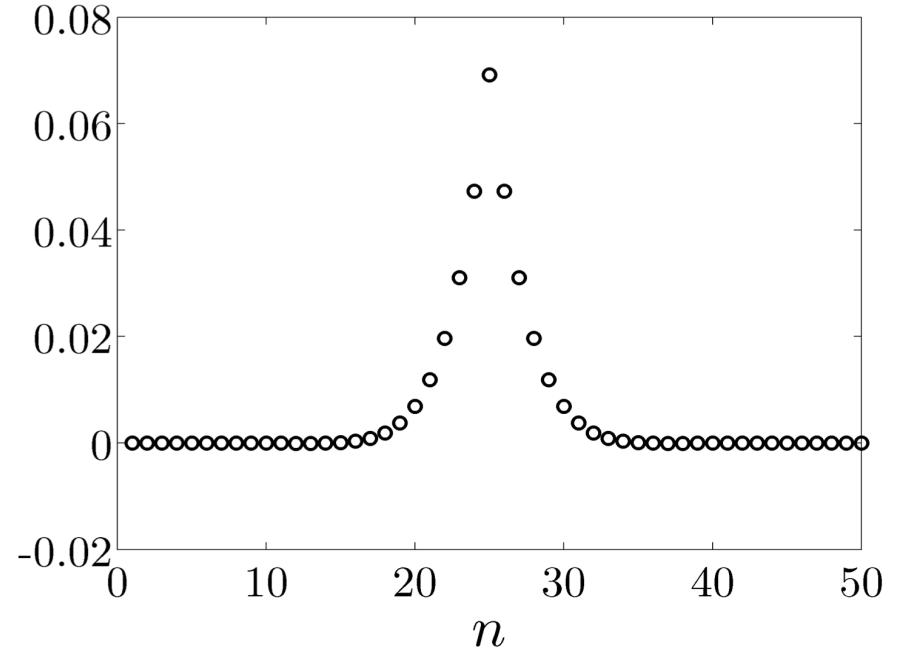
$$\begin{bmatrix} u_1(t) \\ u_2(t) \\ u_3(t) \\ u_4(t) \end{bmatrix} = - \underbrace{\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}}_{F_p} \begin{bmatrix} p_1(t) \\ p_2(t) \\ p_3(t) \\ p_4(t) \end{bmatrix} - \underbrace{\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}}_{F_v} \begin{bmatrix} v_1(t) \\ v_2(t) \\ v_3(t) \\ v_4(t) \end{bmatrix}$$

Structure of optimal controller

position feedback matrix



gains for middle mass



- **OBSERVATIONS**

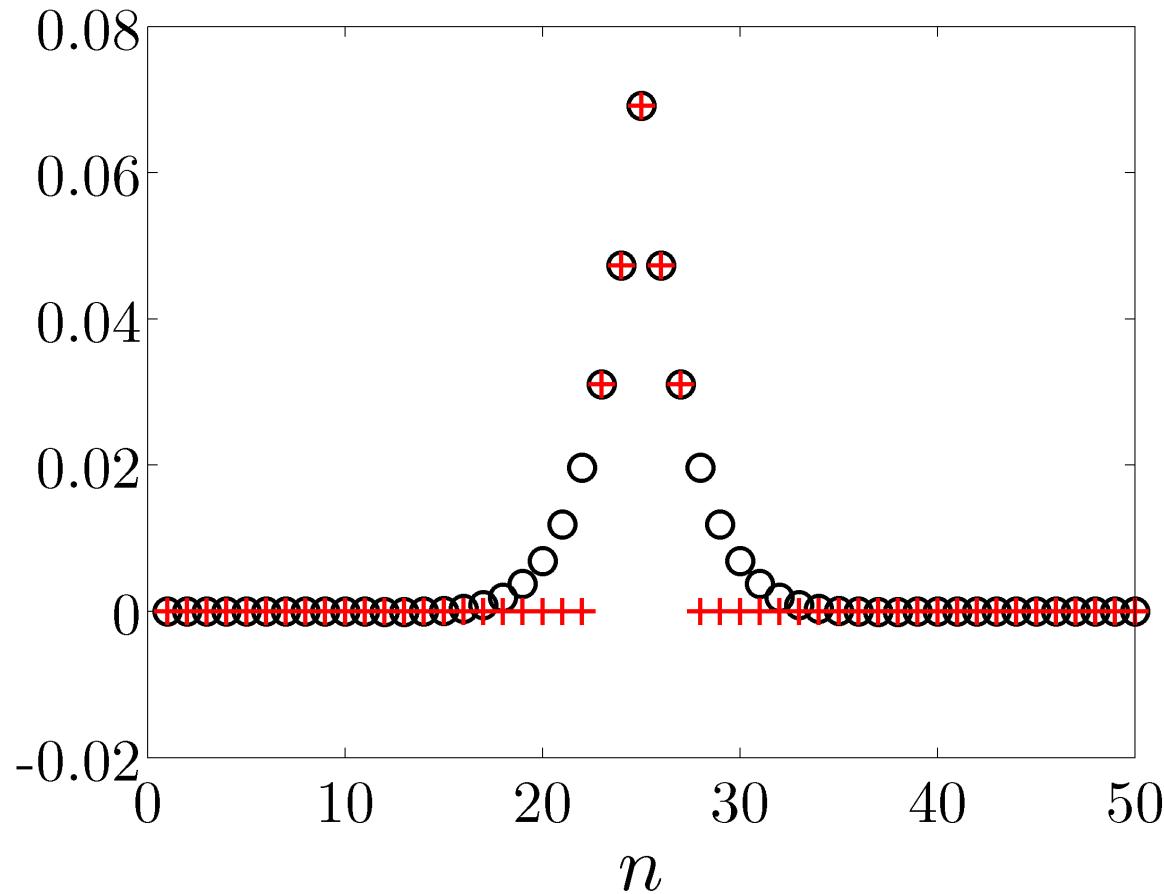
- ★ diagonals almost constant (modulo edges)
- ★ off-diagonal decay of a centralized gain

Bamieh, Paganini, Dahleh, IEEE TAC '02

Mottee & Jadbabaie, IEEE TAC '08

Enforcing sparsity?

- One approach: truncate centralized controller



- DANGERS
 - ★ significant performance degradation
 - ★ instability

Rest of the talk

- SPARSITY-PROMOTING OPTIMAL CONTROL
 - ★ identification and design of sparse feedback gains
- ALGORITHM
 - ★ Proximal Augmented Lagrangian Method
- CLASSES OF CONVEX PROBLEMS
 - ★ optimal actuator/sensor selection
 - ★ optimal design of consensus networks
 - ★ diagonal modifications of positive systems
- EXAMPLES
- SUMMARY AND OUTLOOK

SPARSITY-PROMOTING OPTIMAL CONTROL

- OBJECTIVE

- ★ promote sparsity of F

$$\begin{bmatrix} u_1(t) \\ u_2(t) \\ u_3(t) \\ u_4(t) \\ u_5(t) \end{bmatrix} = - \underbrace{\begin{bmatrix} * & * & & & \\ * & * & & & * \\ & * & * & * & \\ & & * & * & \\ * & & * & * & * \end{bmatrix}}_F \begin{bmatrix} x_1(t) \\ x_2(t) \\ x_3(t) \\ x_4(t) \\ x_5(t) \end{bmatrix}$$

Sparsity-promoting optimal control

$$\text{minimize} \quad J(F) + \gamma \mathbf{card}(F)$$

variance **sparsity-promoting**
amplification **penalty function**

$\mathbf{card}(F)$ – number of non-zero elements of F

$\gamma > 0$ – performance vs sparsity tradeoff

Fardad, Lin, Jovanović, ACC '11

Lin, Fardad, Jovanović, IEEE TAC '13

Convex relaxations of $\text{card}(F)$

ℓ_1 norm: $\sum_{i,j} |F_{ij}|$

weighted ℓ_1 norm: $\sum_{i,j} w_{ij} |F_{ij}|, \quad w_{ij} \geq 0$

- CARDINALITY VS WEIGHTED ℓ_1 NORM

$$\{w_{ij} = 1/|F_{ij}|, \quad F_{ij} \neq 0\} \quad \Rightarrow \quad \text{card}(F) = \sum_{i,j} w_{ij} |F_{ij}|$$

Convex relaxations of $\text{card}(F)$

ℓ_1 norm: $|F_{ij}|$
 i, j

weighted ℓ_1 norm: $w_{ij} |F_{ij}|, \quad w_{ij} \geq 0$
 i, j

- CARDINALITY VS WEIGHTED ℓ_1 NORM

$$\{w_{ij} = 1/|F_{ij}|, \quad F_{ij} = 0\} \Rightarrow \text{card}(F) = \sum_{i,j} w_{ij} |F_{ij}|$$

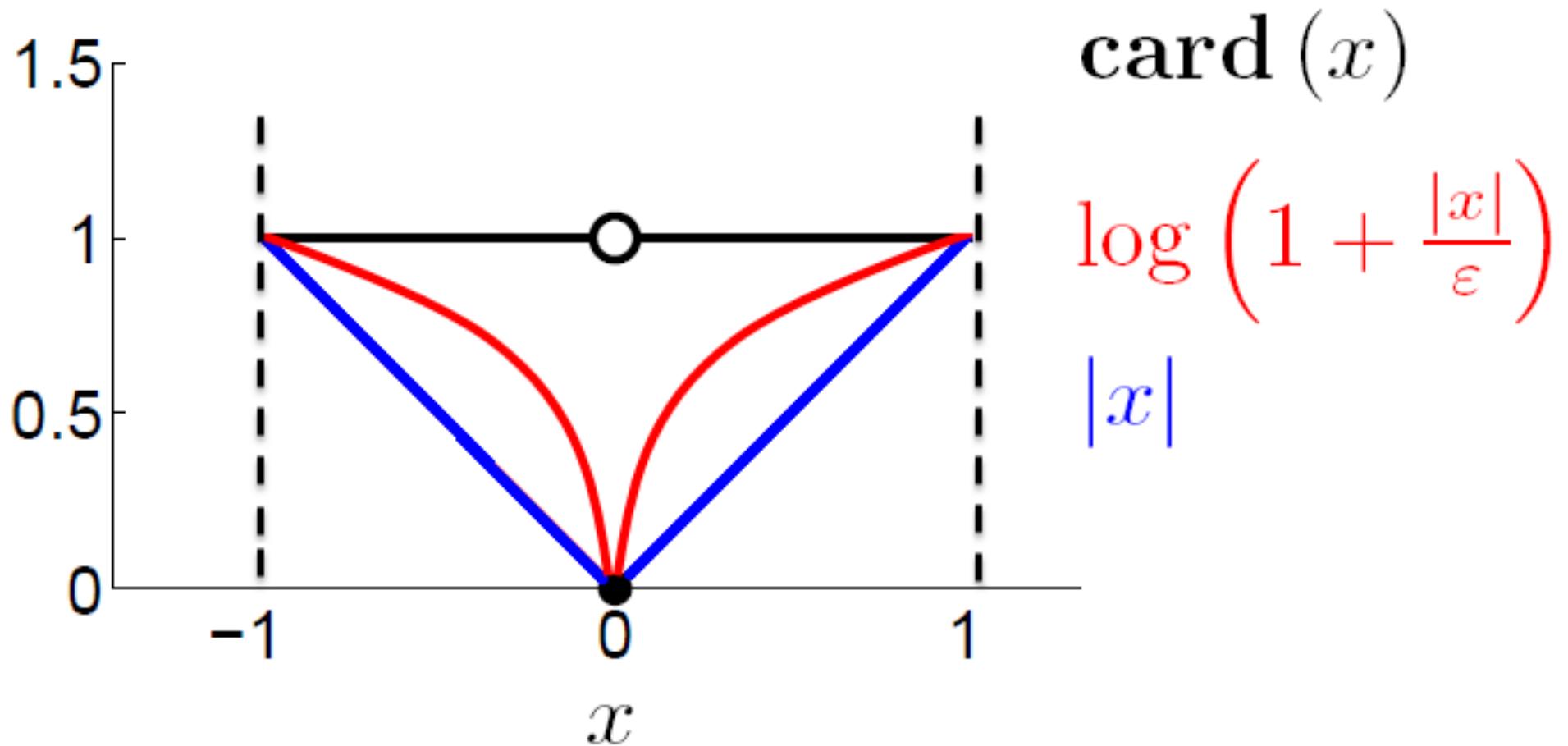
RE-WEIGHTED SCHEME

- ★ use gains from previous iteration to form weights

$$w_{ij}^+ = \frac{1}{|F_{ij}| + \varepsilon}$$

A non-convex relaxation of $\text{card}(F)$

sum-of-logs: $\sum_{i,j} \log \left(1 + \frac{|F_{ij}|}{\varepsilon} \right), \quad 0 < \varepsilon \ll 1$

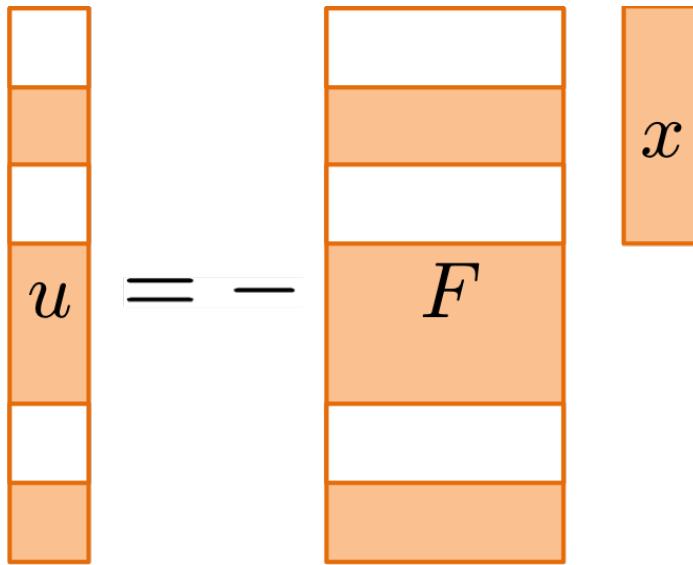


Candès, Wakin, Boyd, J. Fourier Anal. Appl. '08

CLASSES OF CONVEX PROBLEMS

Optimal actuator/sensor selection

- OBJECTIVE: identify **row-sparse** feedback gain



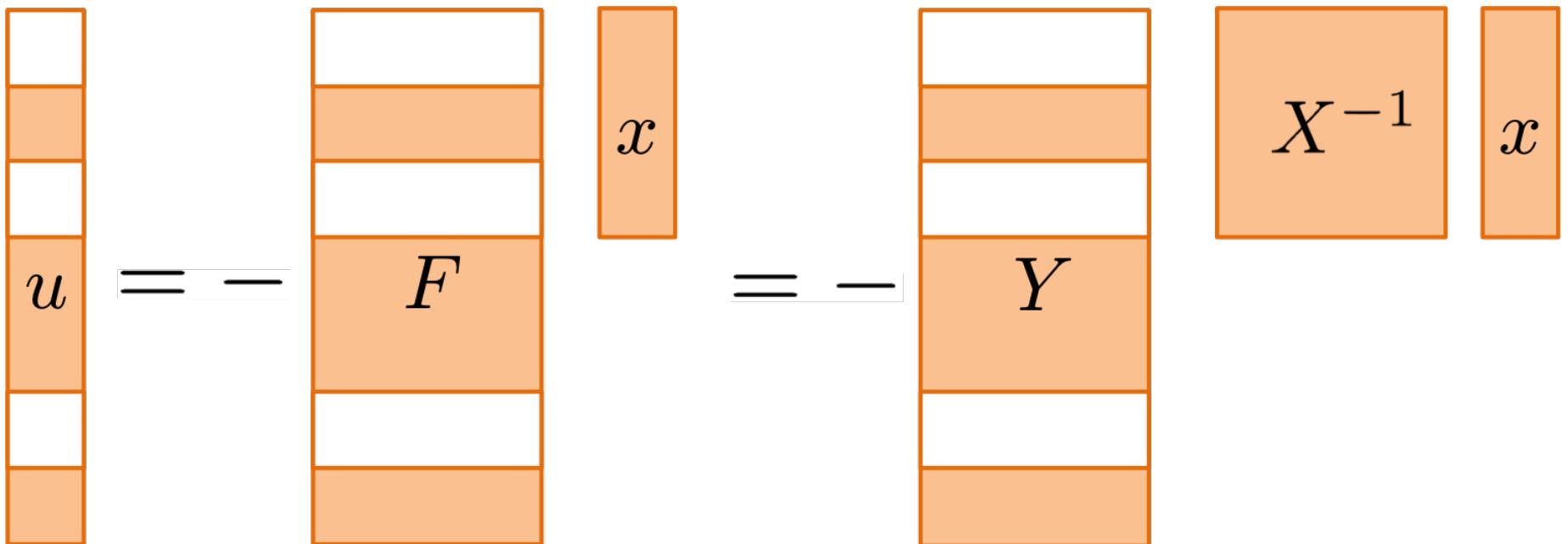
$$\text{minimize} \quad J(F) + \gamma \|e_i^T F\|_2$$

\downarrow i \downarrow
variance amplification **row-sparsity-promoting**
amplification **penalty function**

- CHANGE OF VARIABLES: $Y := F X$

- * convex dependence of J on X and Y

- * row-sparse structure preserved



- OPTIMAL ACTUATOR SELECTION

- admits SDP characterization

$$\text{minimize} \quad J(X, Y) + \gamma \|e_i^T Y\|_2$$

↓ ↓
variance amplification **row-sparsity-promoting
penalty function**

Polyak, Khlebnikov, Shcherbakov, ECC '13

Münz, Pfister, Wolfrum, IEEE TAC '14

Dhingra, Jovanović, Luo, CDC '14

Design of undirected consensus networks

dynamics: $\dot{x} = -\textcolor{red}{L}x + d + u$

control: $u = -\textcolor{blue}{F}x$

objective: $J = \lim_{t \rightarrow \infty} \mathbf{E} (x^T(t) \textcolor{red}{Q} x(t) + u^T(t) \textcolor{red}{R} u(t))$

Design of undirected consensus networks

dynamics: $\dot{x} = -\textcolor{red}{L}x + d + u$

control: $u = -\textcolor{blue}{F}x$

objective: $J = \lim_{t \rightarrow \infty} \mathbf{E} (x^T(t) \textcolor{red}{Q} x(t) + u^T(t) \textcolor{red}{R} u(t))$

convex characterization

minimize $\text{trace}(\textcolor{blue}{X}) + \gamma \mathbb{1}^T \textcolor{blue}{Y} \mathbb{1}$

subject to $\begin{bmatrix} \textcolor{blue}{X} & \begin{bmatrix} Q^{1/2} \\ -R^{1/2} \textcolor{blue}{F} \end{bmatrix} \\ Q^{1/2} - \textcolor{blue}{F} R^{1/2} & \textcolor{blue}{F} + L + \mathbb{1}\mathbb{1}^T/n \end{bmatrix} \succeq 0$

$$\textcolor{blue}{F}\mathbb{1} = 0, \quad -Y_{ij} \leq W_{ij} F_{ij} \leq Y_{ij}$$

Lin, Fardad, Jovanović, Allerton '12

Zelazo, Schuler, Allgöwer, SCL '13

Hassan-Moghaddam & Jovanović, arXiv:1506.03437

Diagonal modifications of positive systems

$$\dot{x} = \left(A + \sum_k u_k D_k \right) x + d$$

A – Metzler matrix ($A_{ij} \geq 0, i = j$)

D_k – diagonal matrices

Diagonal modifications of positive systems

$$\dot{x} = A + \sum_k u_k D_k x + d$$

A – Metzler matrix ($A_{ij} \geq 0, i = j$)

D_k – diagonal matrices

- EXAMPLES

- ★ combination drug therapy

x_i mutates to x_j at rate A_{ji}

u_k kills x_i at rate $(D_k)_{ii}$

- ★ leader selection in directed networks

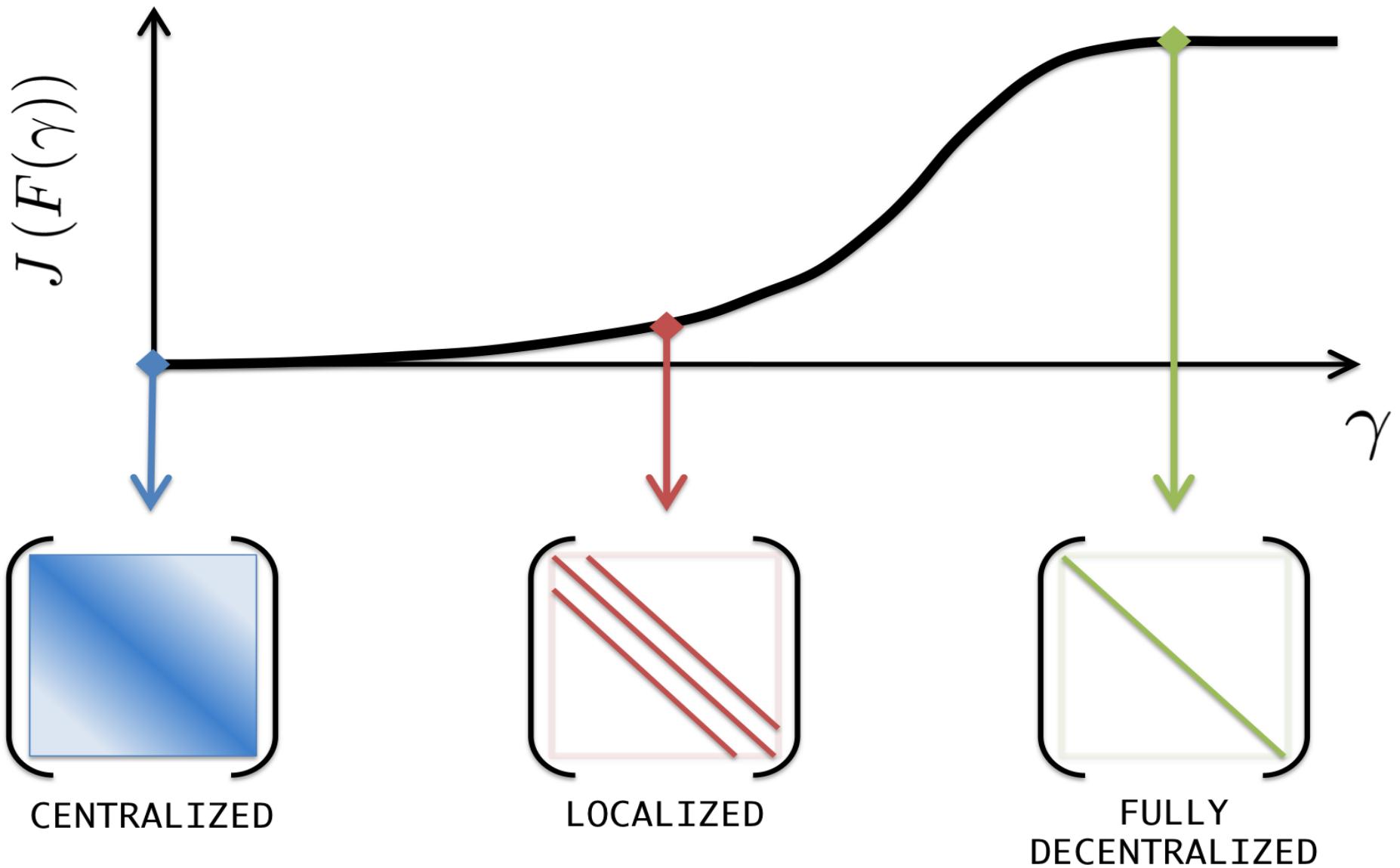
Rantzer & Bernhardsson, CDC '14

Jonsson, Matni, Murray, CDC '14

Dhingra, Colombino, Jovanović, ECC '16

Parameterized family of feedback gains

$$F(\gamma) := \operatorname{argmin}_F (J(F) + \gamma g(F))$$



CASE STUDY: WIDE-AREA CONTROL

Dörfler, Jovanović, Chertkov, Bullo, IEEE TPWRS '14

Wu, Dörfler, Jovanović, IEEE TPWRS '16

<http://people.ece.umn.edu/users/mihailo/software/lqrsp/wac.html>

Electro-mechanical oscillations in power systems

- **Local oscillations**

- single generators swing relative to the rest of the grid
- typically damped by Power System Stabilizers (PSSs)

- **Inter-area oscillations**

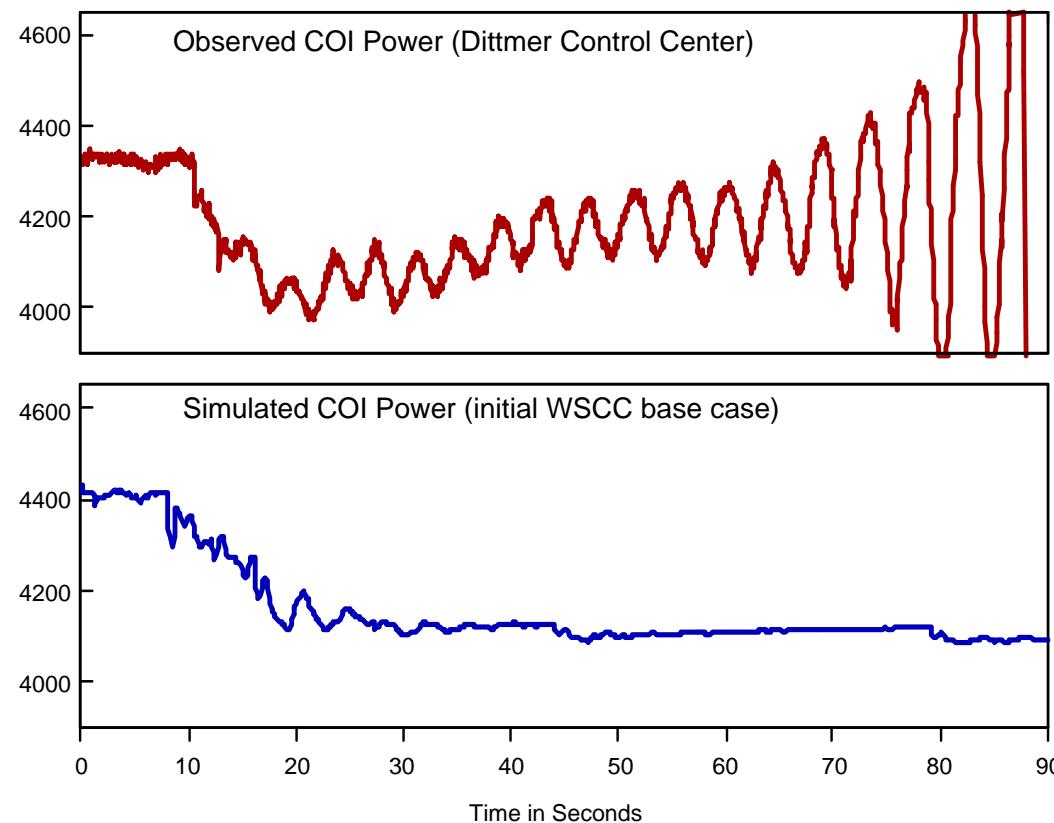
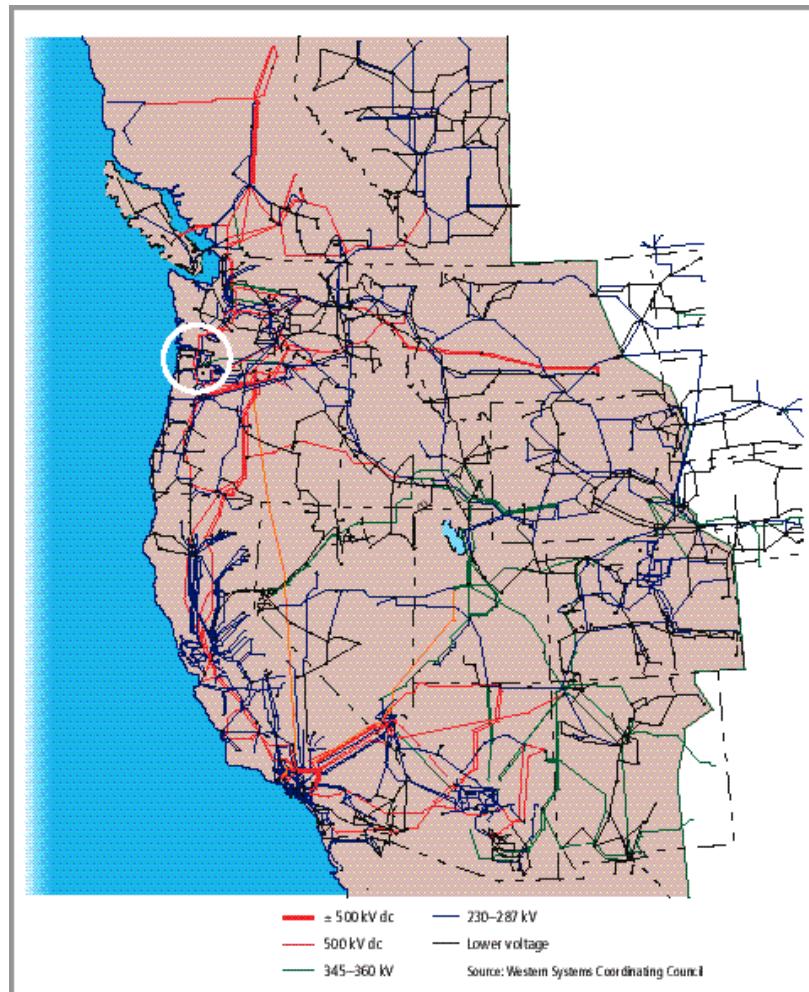
- groups of generators oscillate relative to each other
- associated with dynamics of power transfers

Inter-area oscillations

- Blackout of Aug. 10, 1996
 - ★ resulted from instability of the 0.25 Hz mode

western interconnected system:

California-Oregon power transfer:



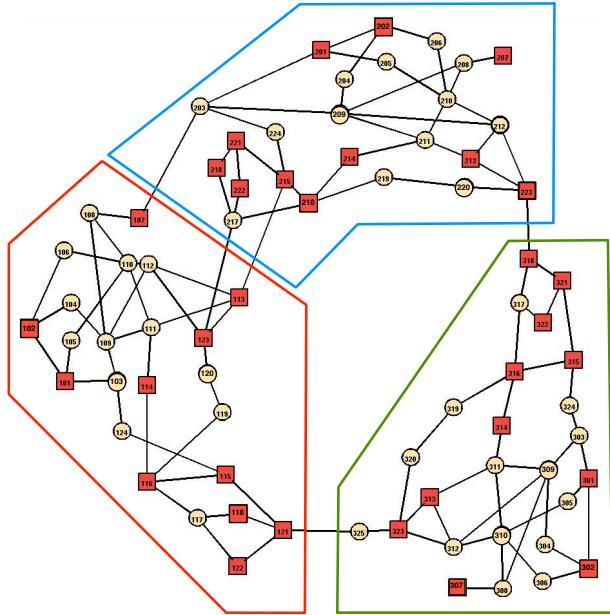
Slow coherency theory

- WHERE ARE THE INTER-AREA MODES COMING FROM?

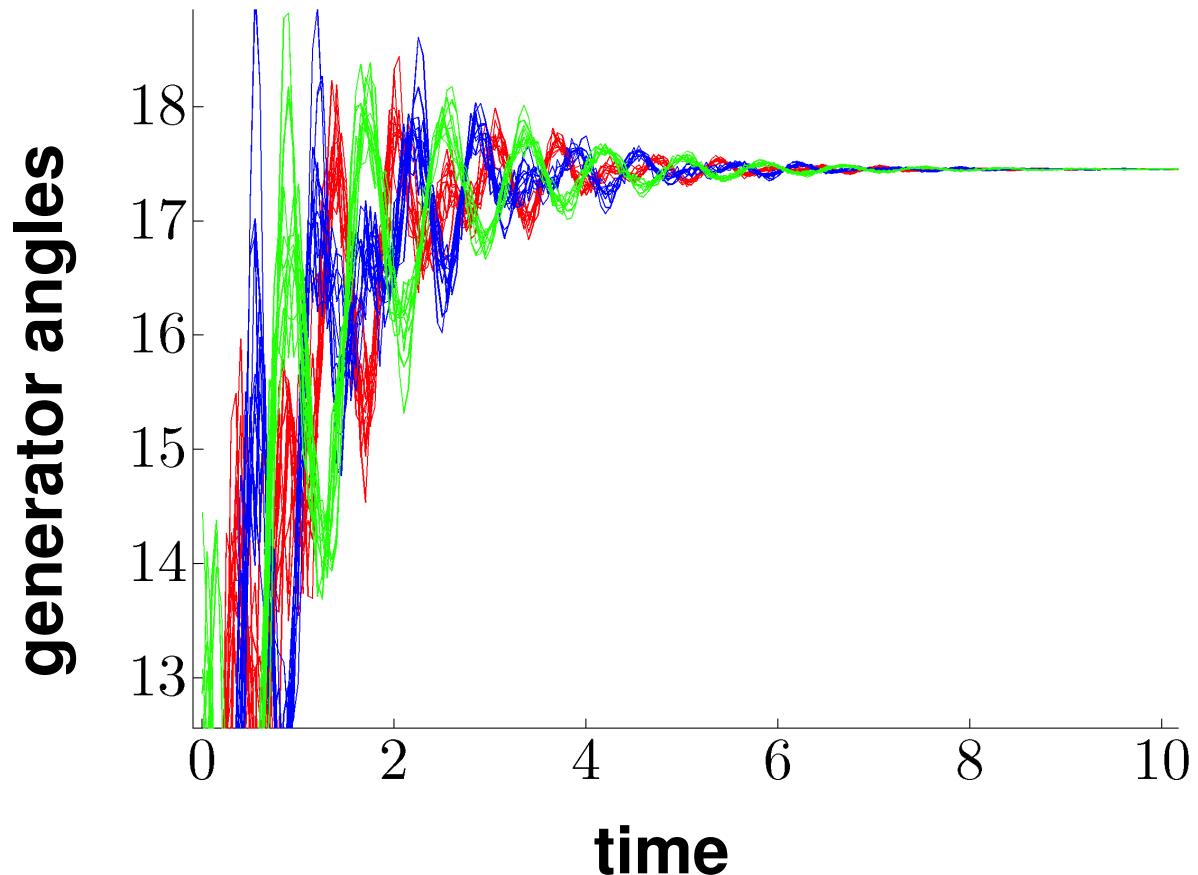
★ slow coherency theory

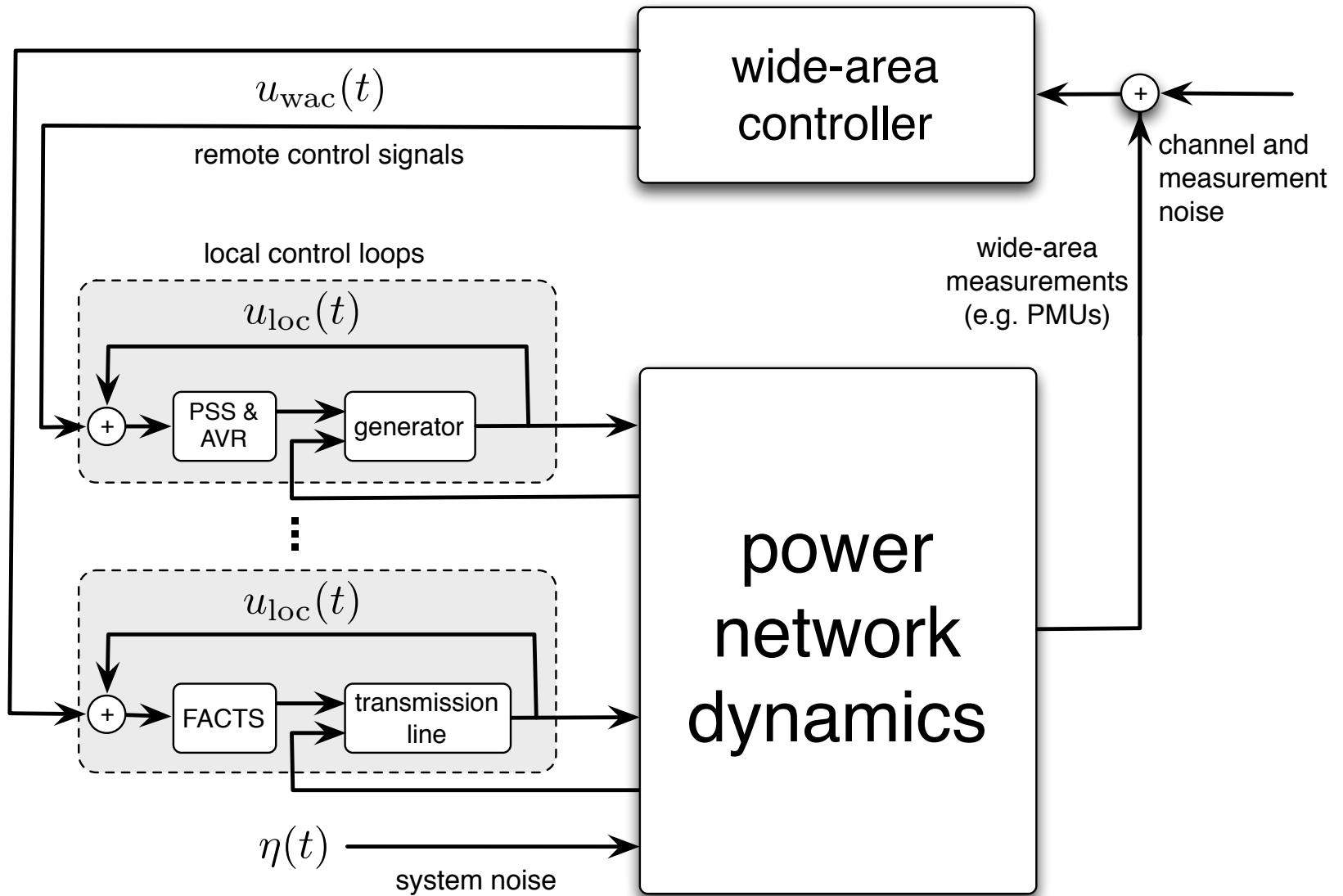
Chow, Kokotović, et al. '78, '82

RTS 96 power system:



linearized swing equation:

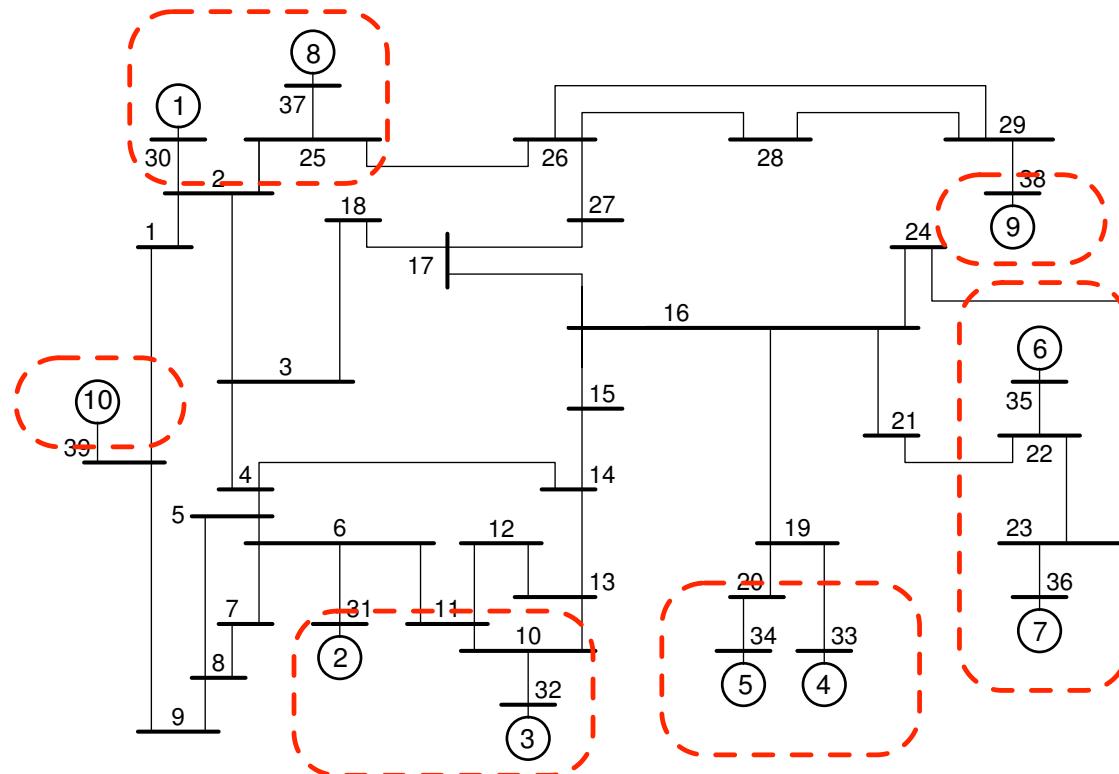




Case study: IEEE New England Power Grid

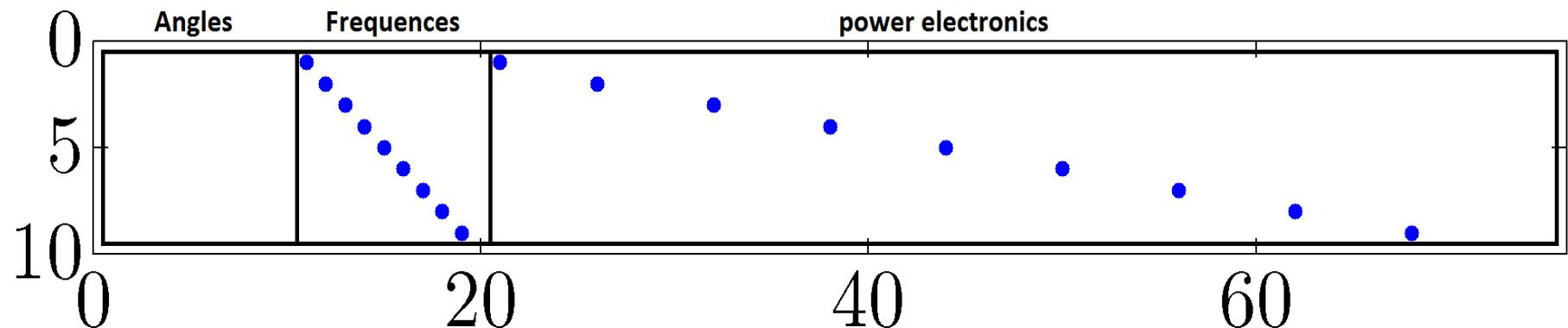
- MODEL FEATURES

- ★ detailed sub-transient generator models
- ★ excitors
- ★ carefully tuned PSS data



Preview of a key result

- FEEDBACK GAIN STRUCTURE



fully decentralized controller

⇒

nearly centralized performance

- ★ 10% degradation relative to the optimal centralized controller
- ★ **optimal retuning** of the decentralized PSS gains

An example: swing equation

$$M \ddot{\theta} + D \dot{\theta} + L \theta = d + u$$

L – Laplacian matrix



only relative angle differences enter into dynamics

Performance index

- ENERGY OF POWER NETWORK

- inspired by slow coherency theory

$$J := \lim_{t \rightarrow \infty} \mathbf{E} \left(\theta^T(t) Q_\theta \theta(t) + \dot{\theta}^T(t) M \dot{\theta}(t) + u^T(t) u(t) \right)$$

$$Q_\theta := I - (1/N) \mathbf{1} \mathbf{1}^T$$

- Q_θ – penalizes deviation from average

$$\bar{\theta} := (1/N) \mathbf{1}^T \theta$$



not detectable from Q_θ

Structural constraints

- ZERO E-VALUE ASSOCIATED WITH THE AVERAGE MODE

open-loop:

$$A \begin{matrix} 1 \\ 0 \end{matrix} = \begin{matrix} 0 \\ 0 \end{matrix}$$

closed-loop: $(A - B_2 K)$

$$\begin{matrix} 1 \\ 0 \end{matrix} = \begin{matrix} 0 \\ 0 \end{matrix}$$

Coordinate transformation

- ELIMINATE THE AVERAGE-MODE

$$\begin{matrix} \theta \\ \dot{\theta} \end{matrix} = \underbrace{\begin{matrix} U & 0 \\ 0 & I \end{matrix}}_T \begin{matrix} \psi \\ \dot{\theta} \end{matrix} + \begin{matrix} 1 \\ 0 \end{matrix} \bar{\theta}$$

BY PROJECTING STATES ONTO

$$\begin{matrix} 1 \\ 0 \end{matrix}^\perp$$

columns of U – form an **orthonormal basis** of $\mathbb{1}^\perp$

Sparsity-promoting optimal control

minimize

$$J(F) + \gamma \|FT^T\|_1$$



new coordinates

(nonconvex, smooth)

original coordinates

(convex, nonsmooth)

- ★ $F = KT$ – to eliminate the average-mode
- ★ $\|FT^T\|_1$ – **not separable** in the elements of F

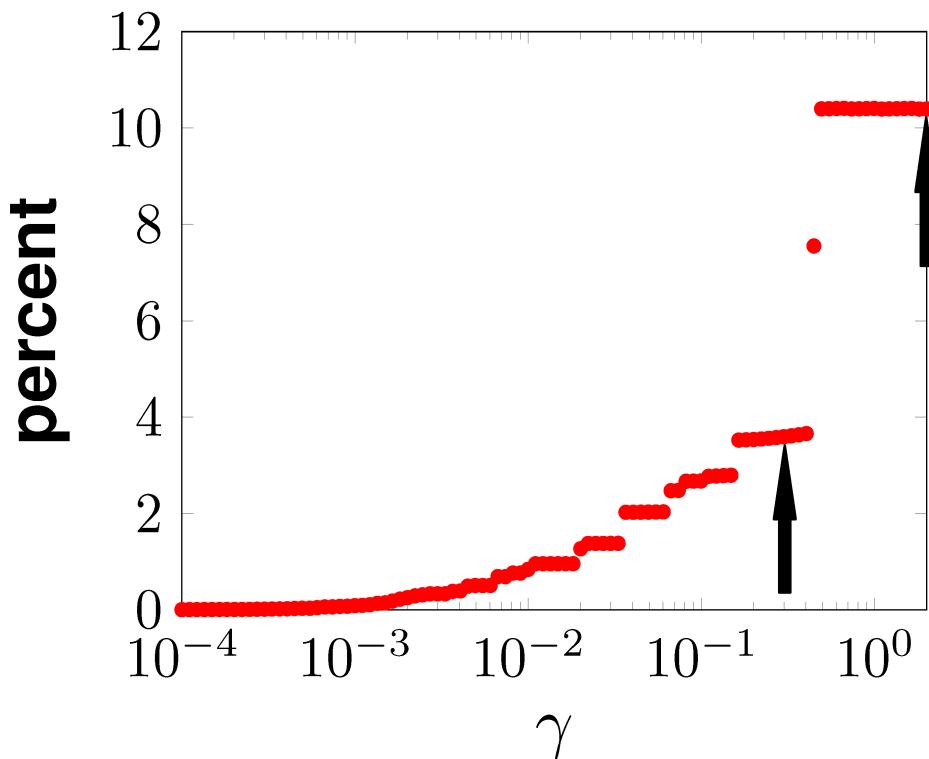
• OPTIMAL CONTROL PROBLEM

$$\underset{F, K}{\text{minimize}} \quad J(\textcolor{blue}{F}) + \gamma \| \textcolor{red}{K} \|_1$$

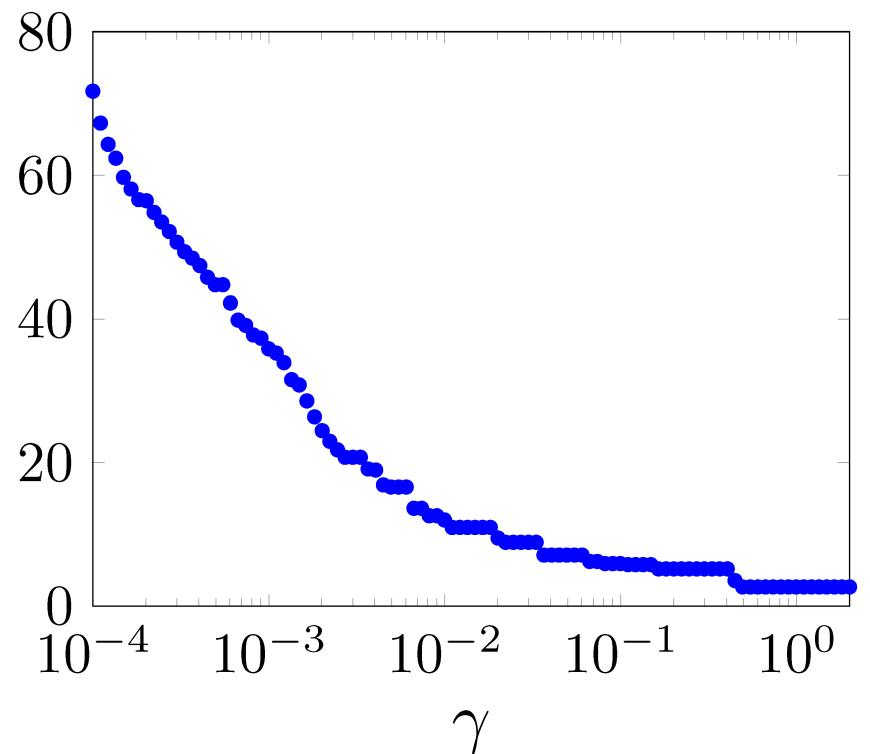
$$\text{subject to} \quad \textcolor{blue}{F} T^T - \textcolor{red}{K} = 0$$

Performance vs sparsity

performance loss:



sparsity of K :

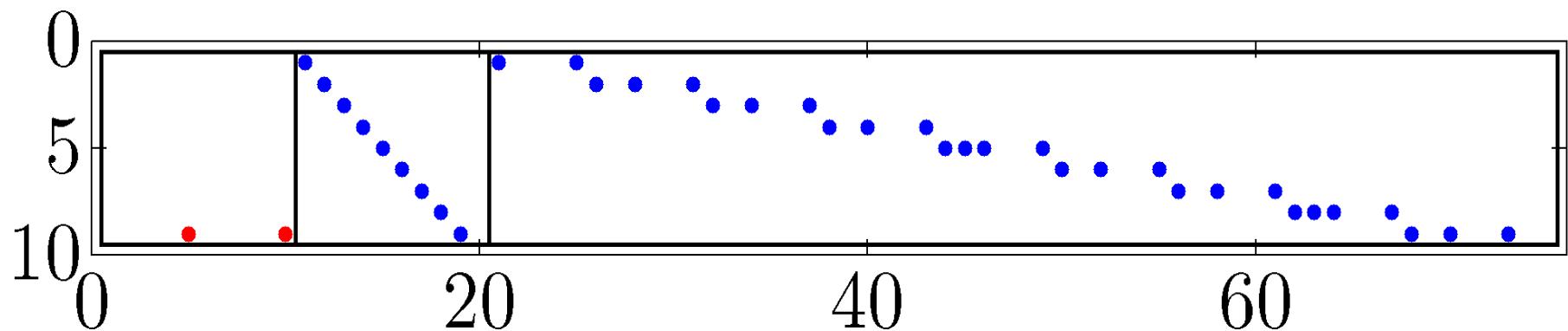


Information exchange network

SPARSITY PATTERN OF K

- local
- long-range interactions

$$\gamma = 0.1099, \text{card}(K) = 39$$

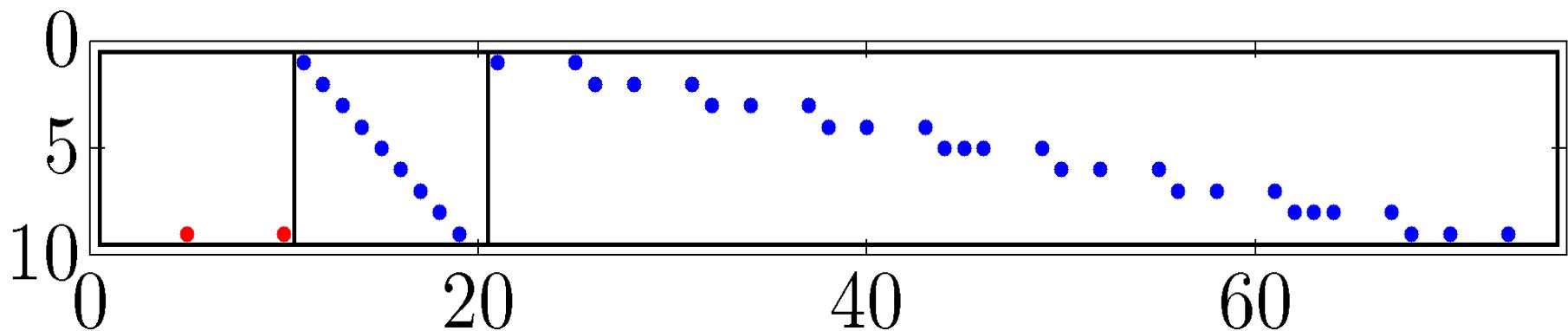


Information exchange network

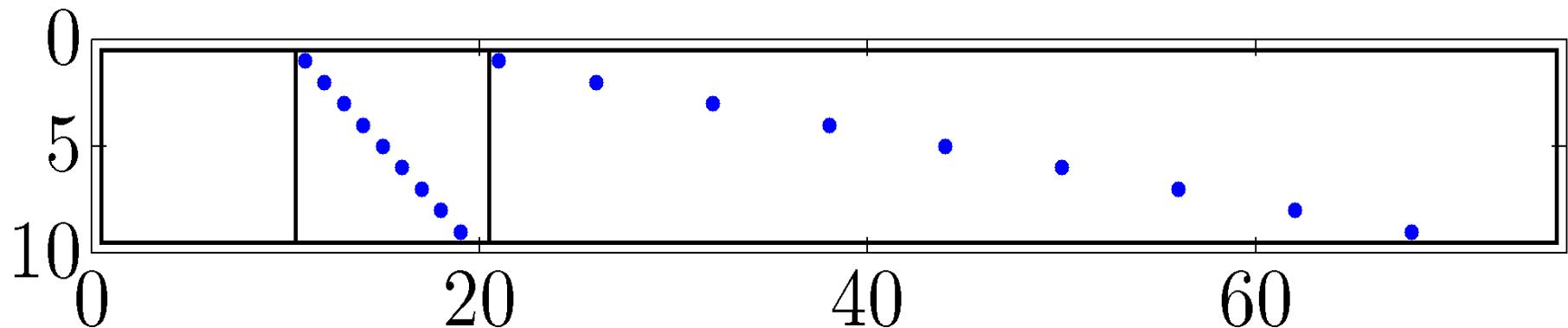
SPARSITY PATTERN OF K

- local
- long-range interactions

$$\gamma = 0.1099, \text{card}(K) = 39$$



$$\gamma = 2, \text{card}(K) = 18$$



Response to stochastic forcing

- WHITE-IN-TIME FORCING

$$\mathbf{E} (d(t_1) d^*(t_2)) = I \delta(t_1 - t_2)$$

- ★ Hilbert-Schmidt norm

power spectral density:

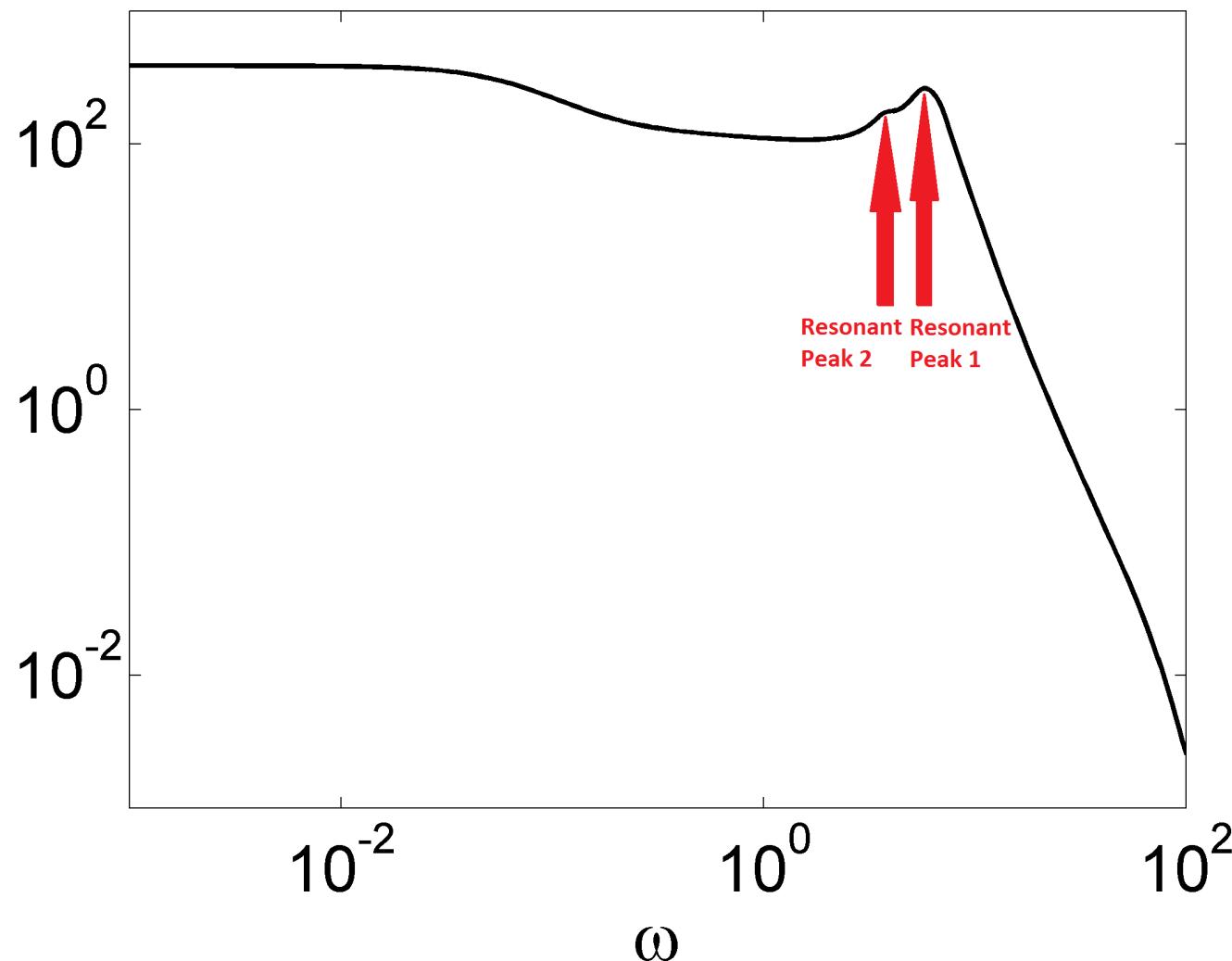
$$\|H(\omega)\|_{\text{HS}}^2 = \text{trace}(H(\omega) H^*(\omega)) = \sum_i \sigma_i^2(\omega)$$

- ★ H_2 norm

variance amplification:

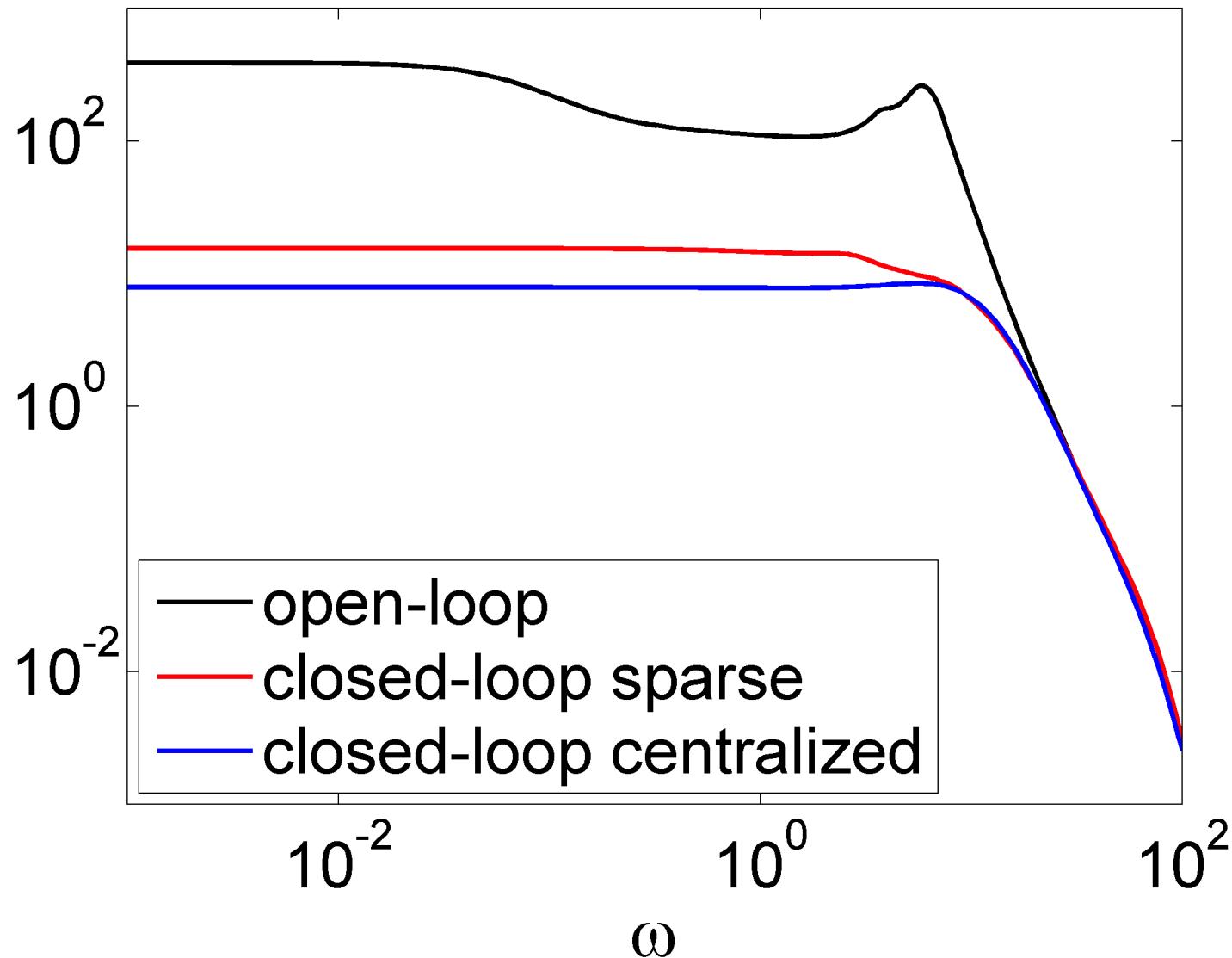
$$\|H\|_2^2 = \frac{1}{2\pi} \int_{-\infty}^{\infty} \|H(\omega)\|_{\text{HS}}^2 d\omega$$

Open-loop dynamics: power spectral density



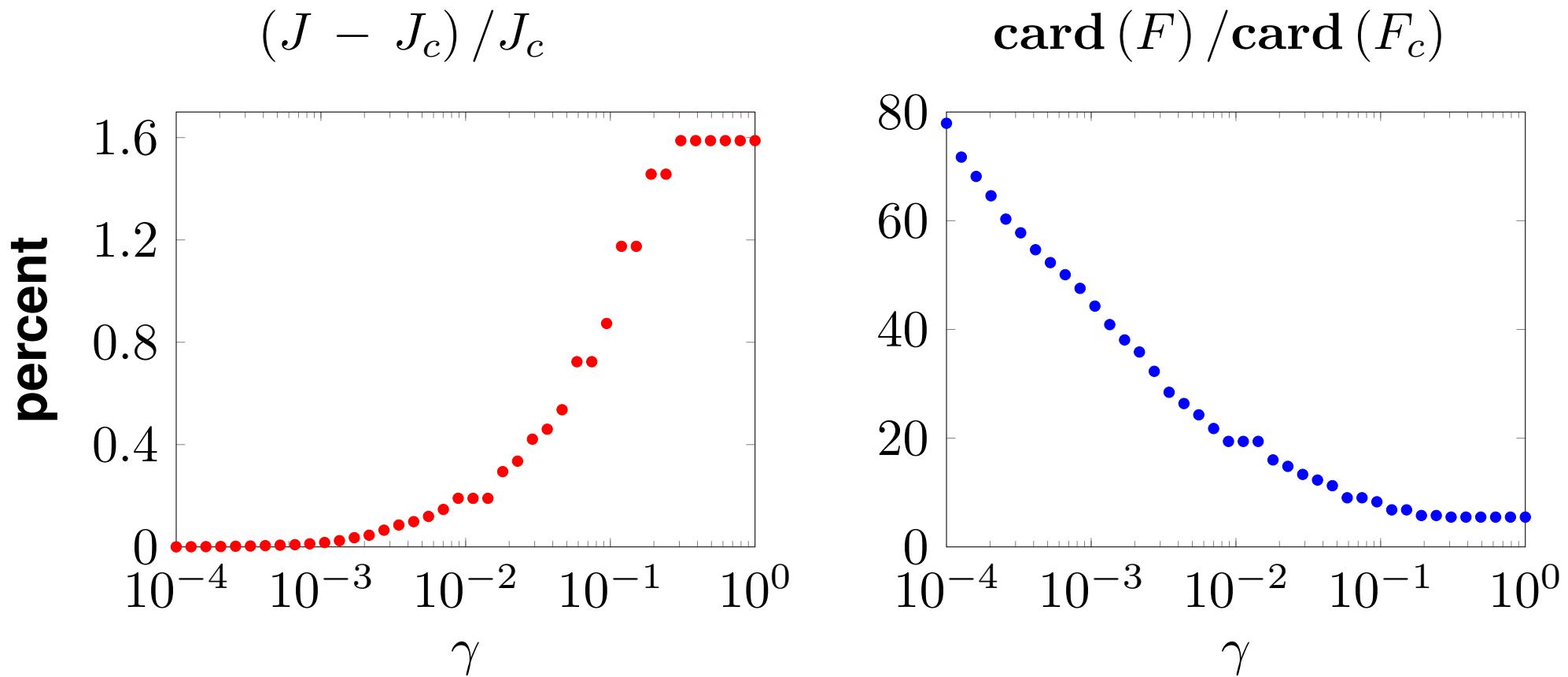
- ★ resonant peak 1: inter-area modes 2, 3, 4, 5
- ★ resonant peak 2: inter-area mode 1

Open-loop vs closed-loop systems



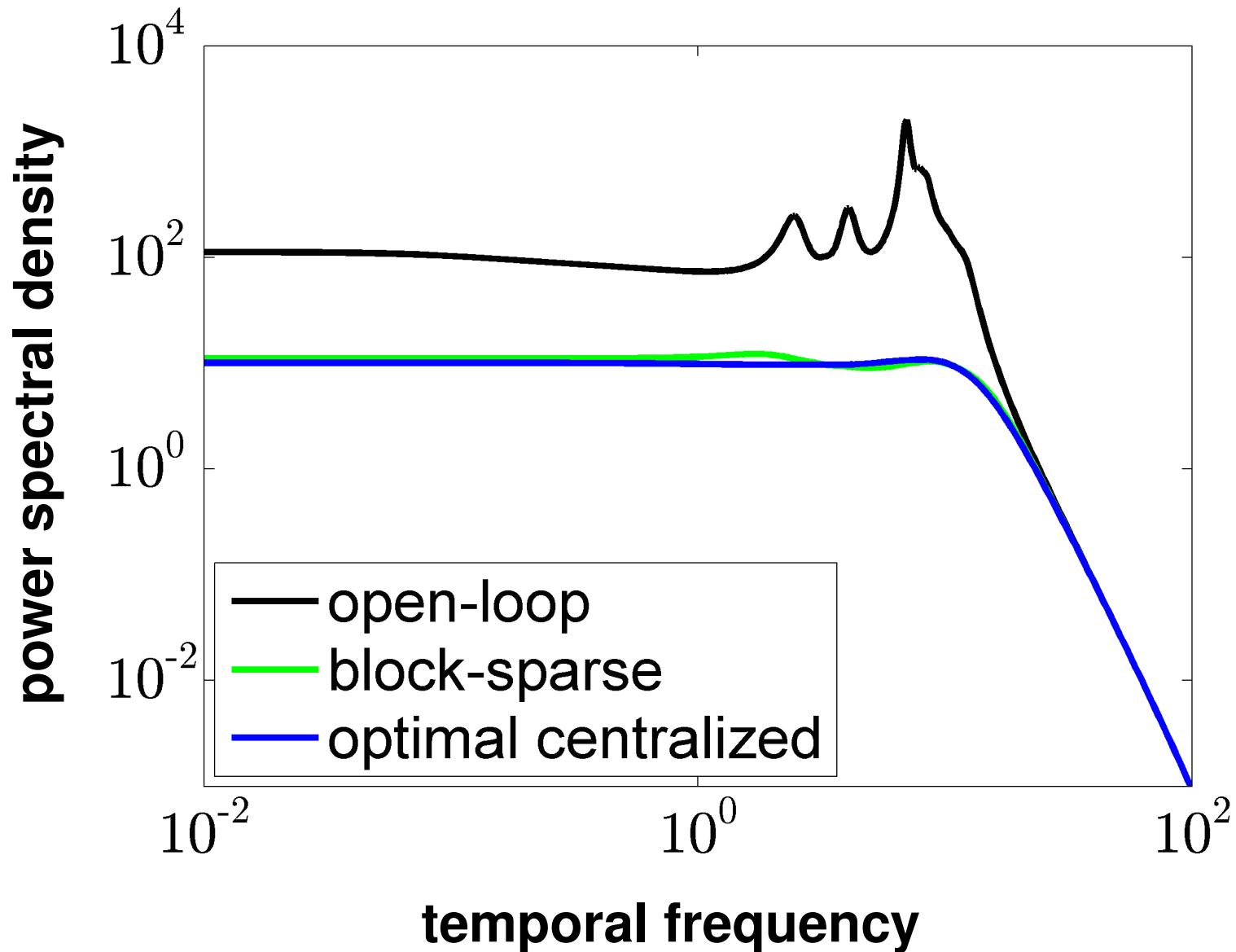
- ★ low frequencies: 10% performance degradation

- Performance comparison: **block-sparse vs centralized**



$\gamma = 1$ $\xrightarrow{\text{relative to } F_c}$ 1.6 % performance loss
5.5 % non-zero elements in F

- RE-DESIGN OF FULLY-DECENTRALIZED CONTROLLERS
 - ★ preserves rotational symmetry

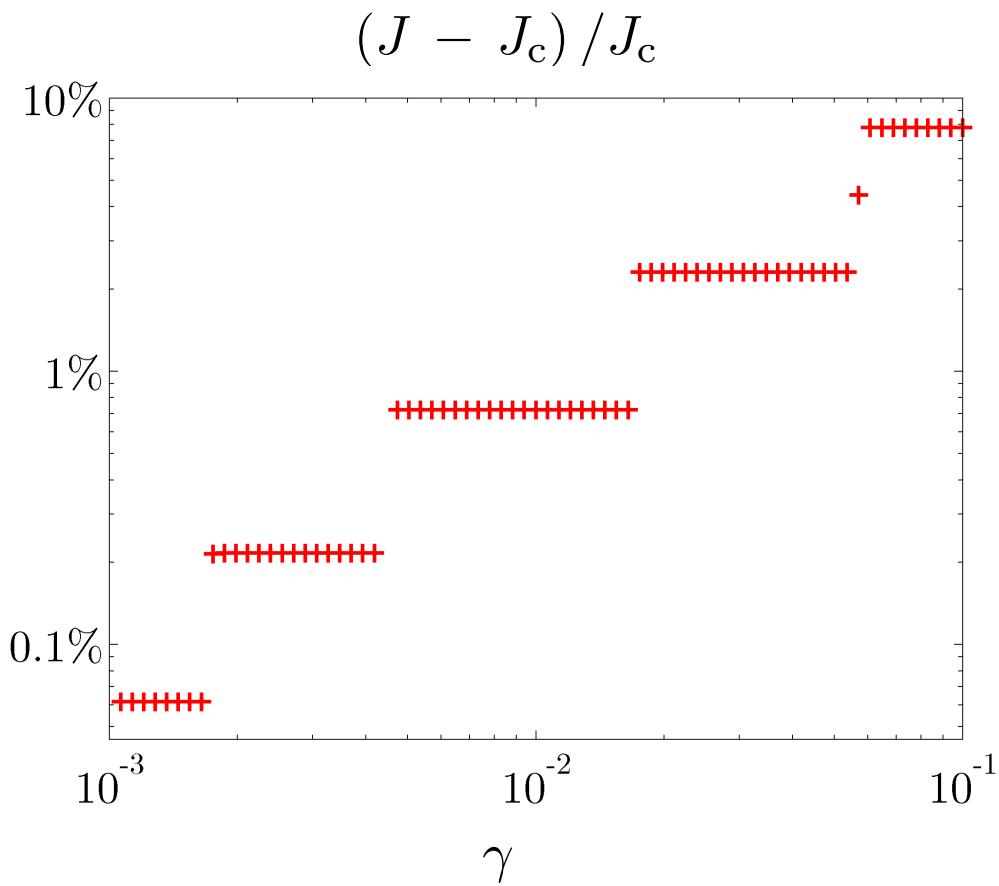
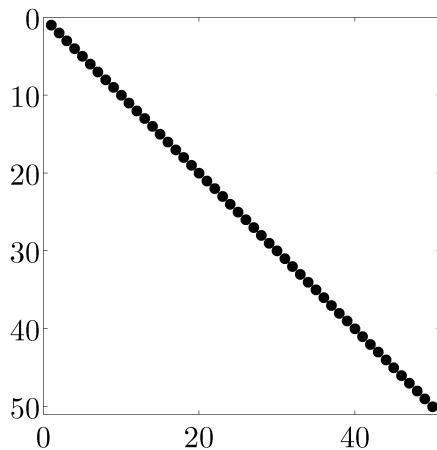
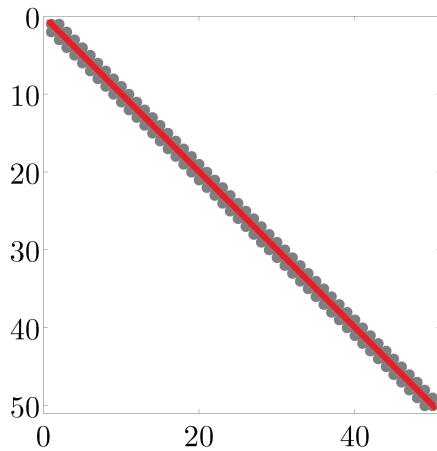
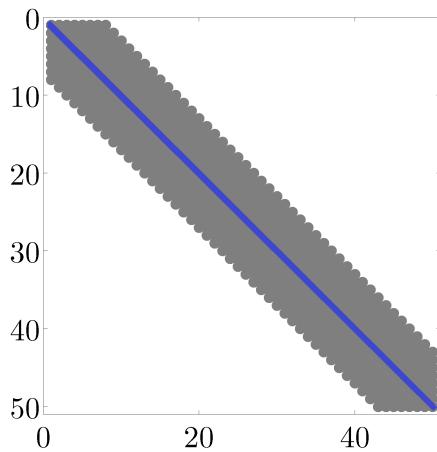


ADDITIONAL EXAMPLES

www.umn.edu/~mihailo/software/lqrsp/

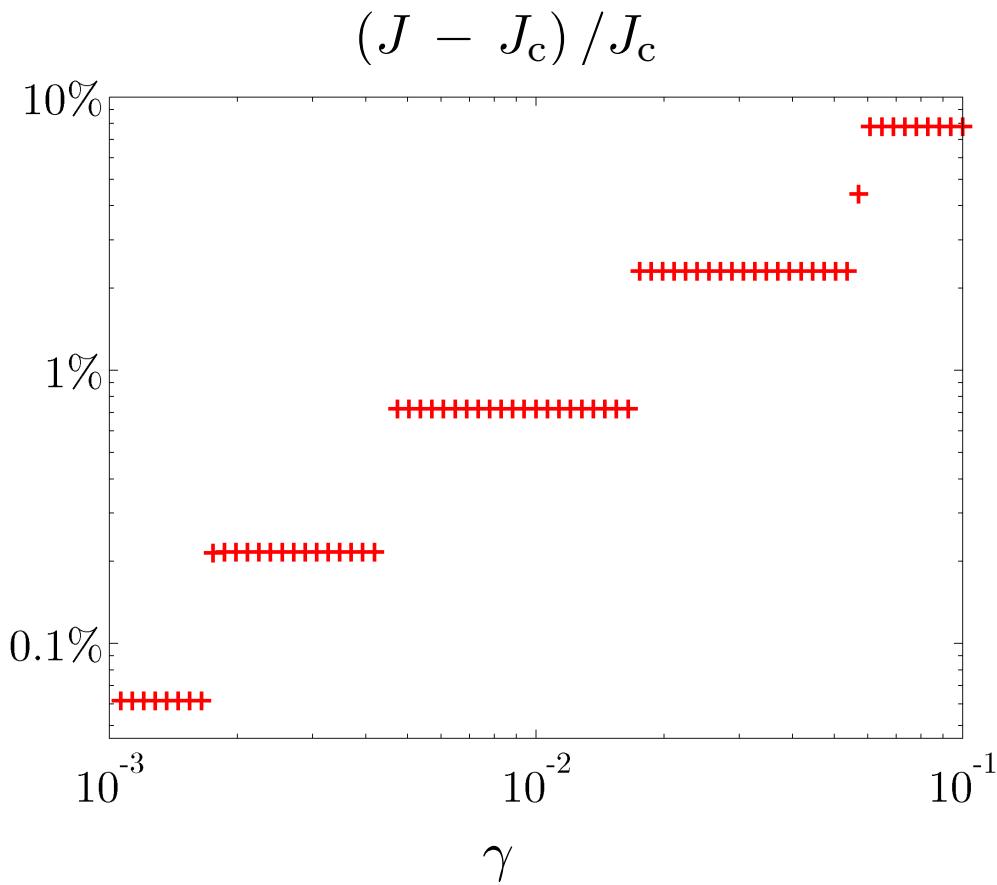
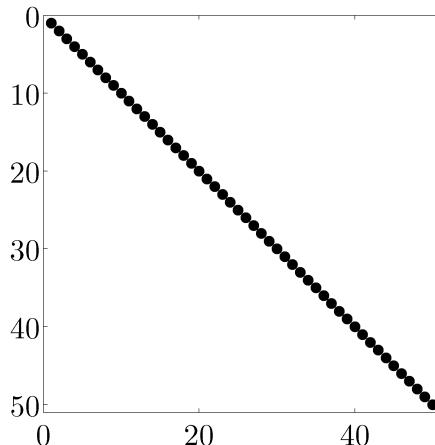
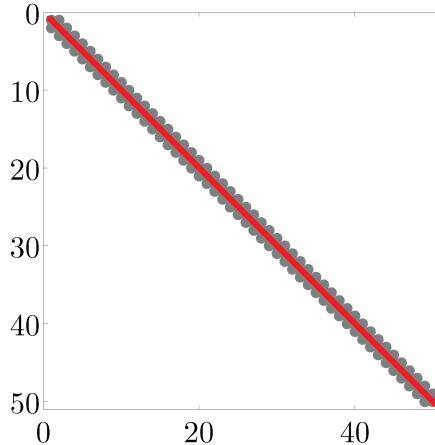
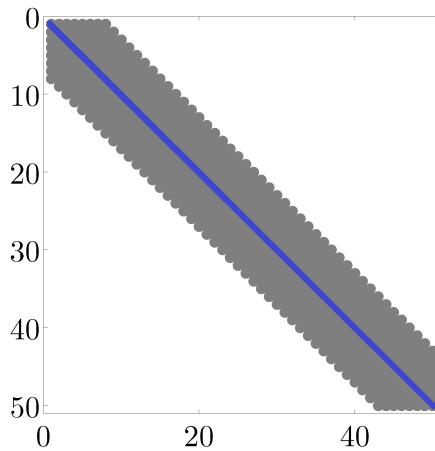
Mass-spring system

Performance comparison: **sparse** vs **centralized**



Mass-spring system

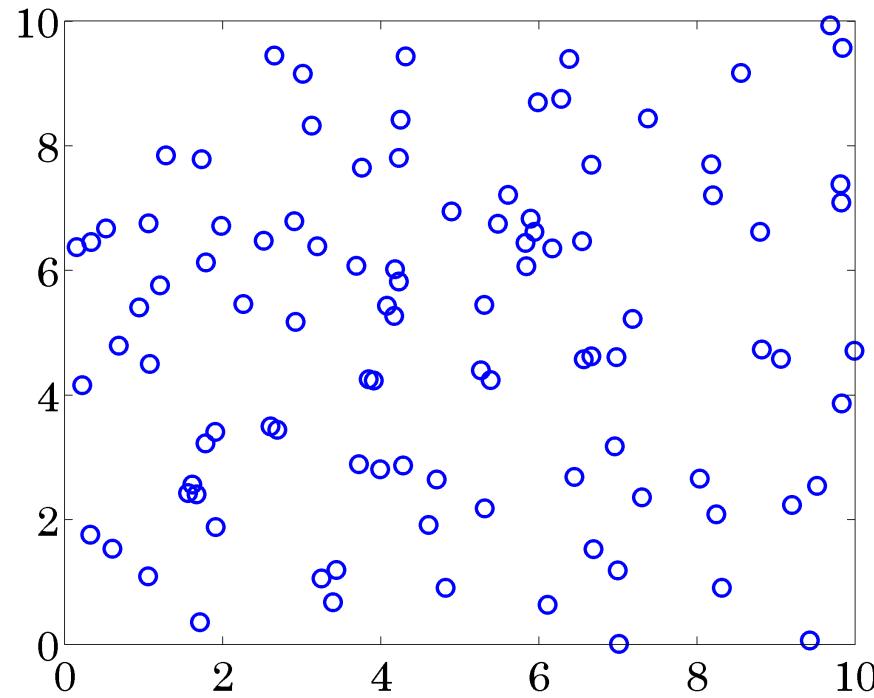
Performance comparison: **sparse vs centralized**



$\text{card}(F) / \text{card}(F_c)$	$(J - J_c) / J_c$
10%	0.75%
6%	2.4%
2%	7.8%

fully-decentralized

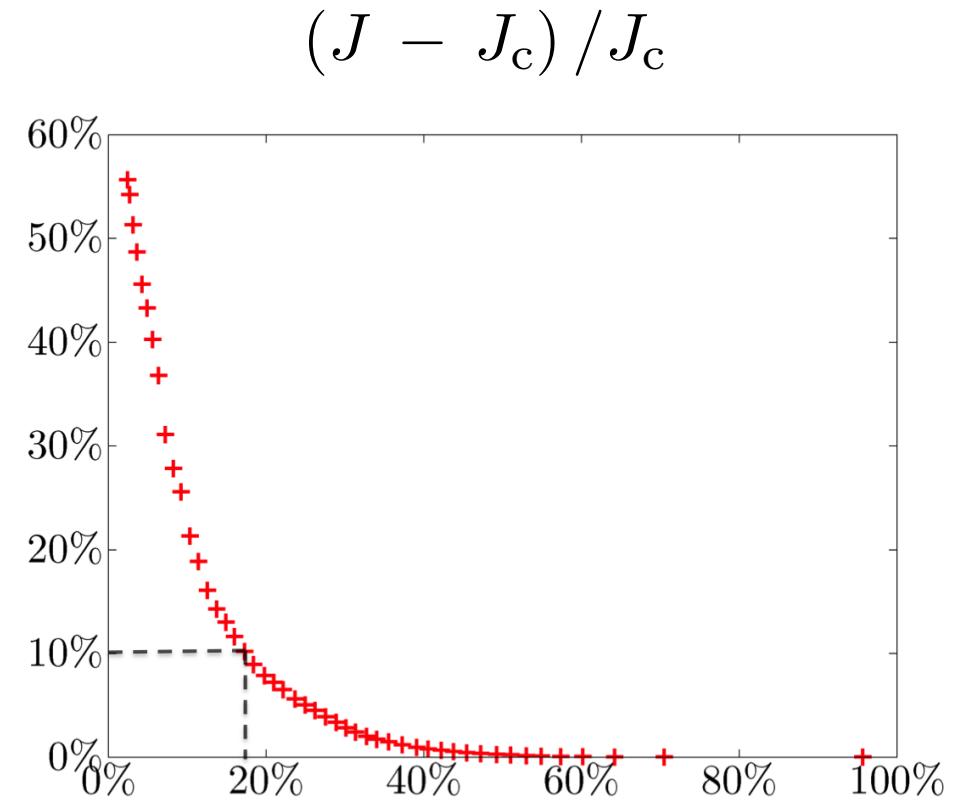
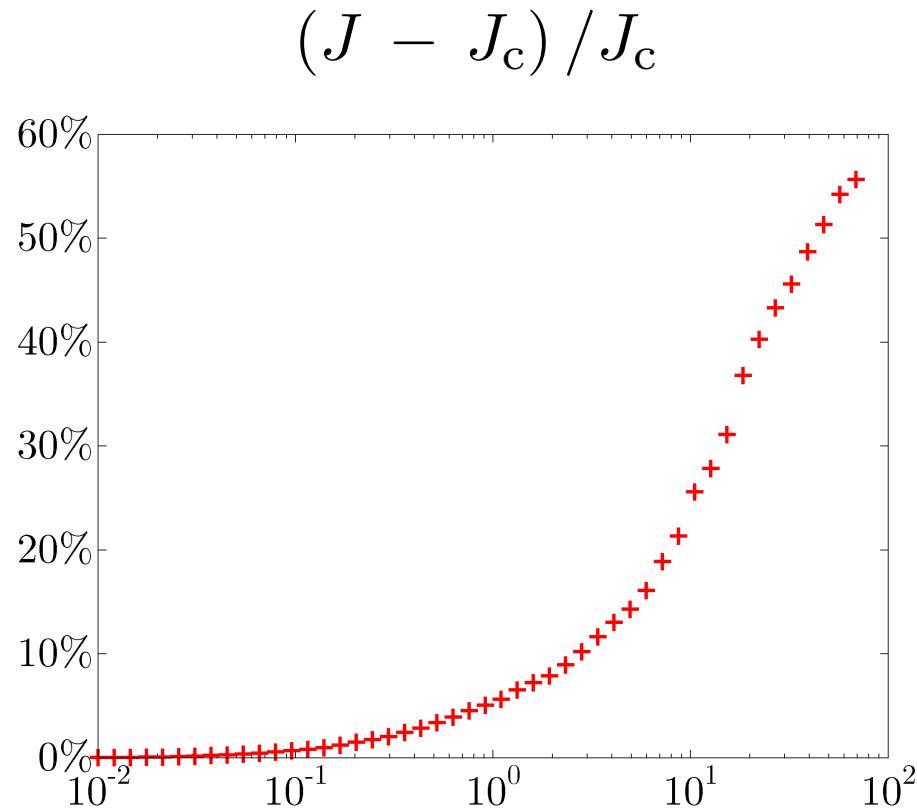
Network with 100 nodes



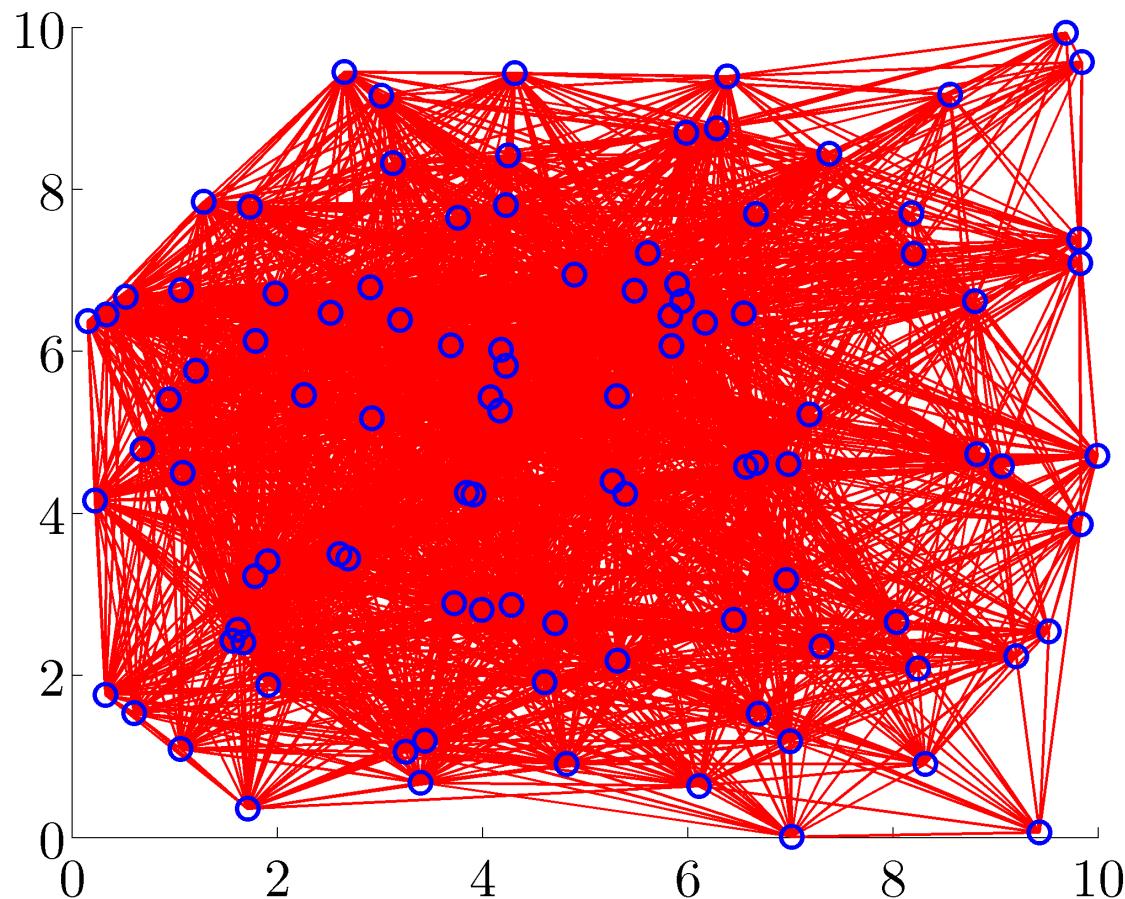
$$\begin{aligned} \dot{p}_i \\ \dot{v}_i \end{aligned} = \underbrace{\begin{pmatrix} 1 & 1 \\ 1 & 2 \end{pmatrix} \begin{pmatrix} p_i \\ v_i \end{pmatrix}}_{\text{unstable dynamics}} + \underbrace{\sum_{j \neq i} e^{-\alpha(i,j)} \begin{pmatrix} p_j \\ v_j \end{pmatrix}}_{\text{coupling}} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} (d_i + u_i)$$

$\alpha(i, j)$: Euclidean distance between nodes i and j

- Performance comparison: **sparse vs centralized**


 γ
 $\text{card}(F) / \text{card}(F_c)$

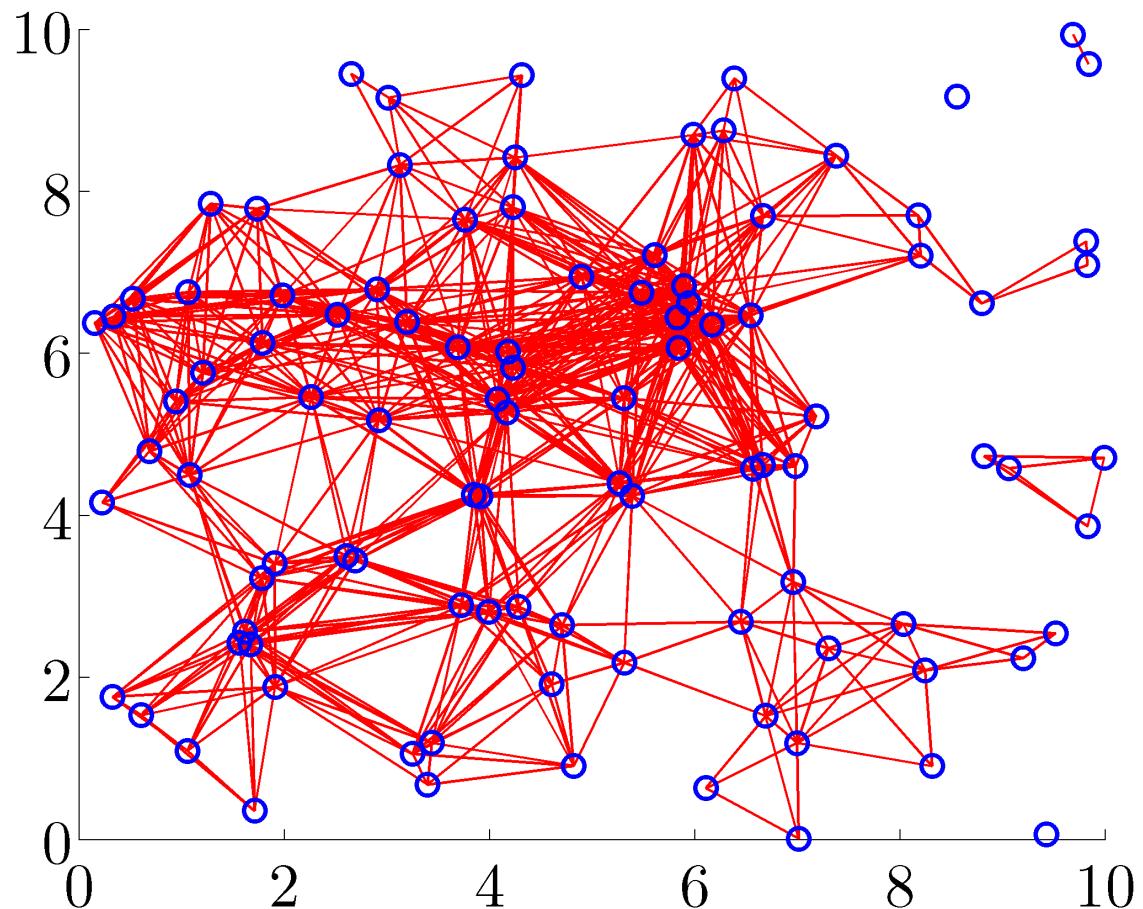
communication graph of a truncated centralized gain



$\text{card}(F) = 7380 \ (36.9\%)$

non-stabilizing

identified communication graph



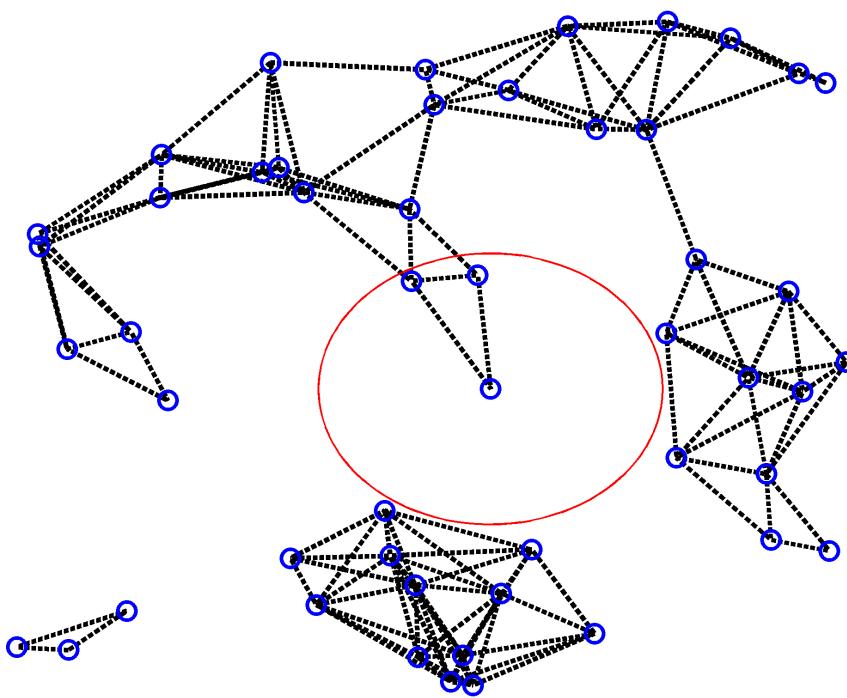
$$\gamma = 5$$

$$\text{card}(F) / \text{card}(F_c) = 8.8\%$$

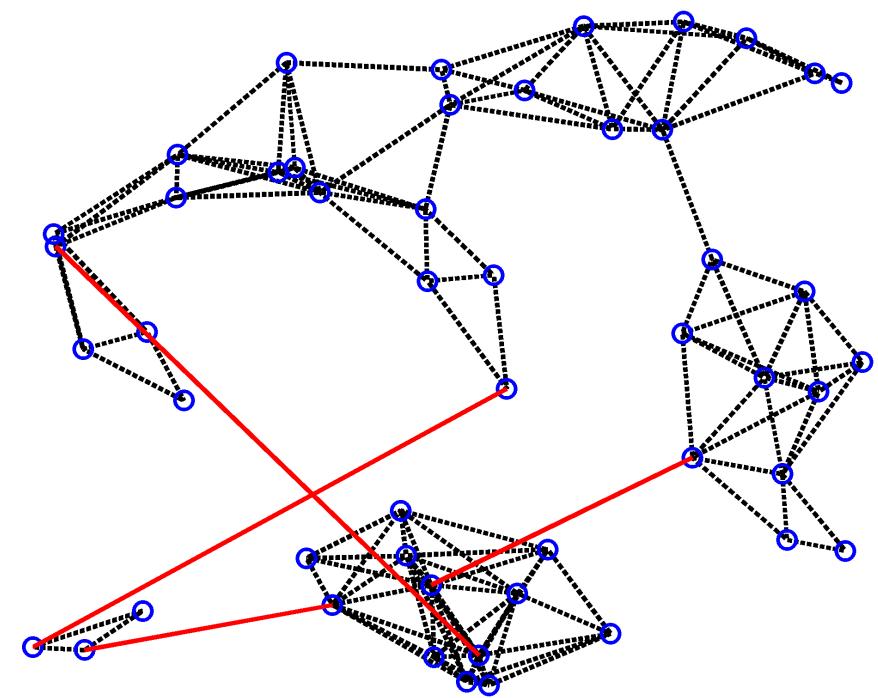
$$(J - J_c) / J_c = 24.6\%$$

Sparsity-promoting consensus algorithm

plant graph



identified communication graph



$Q :=$ deviation from average

$$\frac{J - J_{\text{all-to-all}}}{J_{\text{all-to-all}}} \approx 82\%$$

ALGORITHM

Method of multipliers

$$\text{minimize } J(F) + \gamma g(F)$$

- Step 1: introduce an additional variable/constraint

$$\text{minimize } J(F) + \gamma g(G)$$

$$\text{subject to } F - G = 0$$

benefit: decouples J and g

Method of multipliers

$$\text{minimize } J(F) + \gamma g(F)$$

- Step 1: introduce an additional variable/constraint

$$\text{minimize } J(F) + \gamma g(G)$$

$$\text{subject to } F - G = 0$$

benefit: decouples J and g

- Step 2: introduce augmented Lagrangian

$$\mathcal{L}_\rho(F, G; \Lambda) = J(F) + \gamma g(G) + \langle \Lambda, F - G \rangle + \frac{\rho}{2} \|F - G\|_F^2$$

- **Step 3: use MM for augmented Lagrangian minimization**

$$\mathcal{L}_\rho(F, G; \Lambda) = J(F) + \gamma g(G) + \langle \Lambda, F - G \rangle + \frac{\rho}{2} \|F - G\|_F^2$$

METHOD OF MULTIPLIERS

$$(F^{k+1}, G^{k+1}) := \underset{F, G}{\operatorname{argmin}} \mathcal{L}_{\rho^k}(F, G; \Lambda^k)$$

$$\Lambda^{k+1} := \Lambda^k + \rho^k (F^{k+1} - G^{k+1})$$

- **Step 4: Polishing – back to structured optimal design**

* MM

identifies sparsity patterns

provides good initial condition for structured design

- **Step 4: Polishing – back to structured optimal design**

- ★ **MM**
 - identifies sparsity patterns
 - provides good initial condition for structured design
- ★ **optimality conditions for the structured problem**

$$(A - B_2 \mathbf{F})^T \mathbf{P} + \mathbf{P} (A - B_2 \mathbf{F}) = - (Q + \mathbf{F}^T R \mathbf{F})$$

$$(A - B_2 \mathbf{F}) \mathbf{X} + \mathbf{X} (A - B_2 \mathbf{F})^T = -B_1 B_1^T$$

$$[(R \mathbf{F} - B_2^T \mathbf{P}) \mathbf{X} \circ I_{\mathcal{S}}] = 0$$

$I_{\mathcal{S}}$ - structural identity

$$F = \begin{bmatrix} * & * \\ * & * & * \\ * & * & * \\ * & * \end{bmatrix} \Rightarrow I_{\mathcal{S}} = \begin{bmatrix} 1 & 1 \\ 1 & 1 & 1 \\ 1 & 1 & 1 \\ 1 & 1 \end{bmatrix}$$

Proximal operator and Moreau envelope

- PROXIMAL OPERATOR

$$\text{prox}_{\mu g}(V) := \operatorname{argmin}_G g(G) + \frac{1}{2\mu} \|G - V\|_F^2$$

MOREAU ENVELOPE

$$M_{\mu g}(V) := \inf_G g(G) + \frac{1}{2\mu} \|G - V\|_F^2$$

Proximal operator and Moreau envelope

- PROXIMAL OPERATOR

$$\text{prox}_{\mu g}(V) := \operatorname{argmin}_G g(G) + \frac{1}{2\mu} \|G - V\|_F^2$$

MOREAU ENVELOPE

$$M_{\mu g}(V) := \inf_G g(G) + \frac{1}{2\mu} \|G - V\|_F^2$$

- ★ **continuously differentiable**
even when g is not

$$\nabla M_{\mu g}(V) = \frac{1}{\mu} (V - \text{prox}_{\mu g}(V))$$

Proximal augmented Lagrangian

$$\mathcal{L}_\rho(F, G; \Lambda) = J(F) + \underbrace{\gamma g(G) + \frac{\rho}{2} \|G - (F + (1/\rho)\Lambda)\|_F^2}_{\text{the proximal augmented Lagrangian}} - \frac{1}{2\rho} \|\Lambda\|_F^2$$

★ minimize over G

$$G^\star = \text{prox}_{(\gamma/\rho)g}(F + (1/\rho)\Lambda)$$

★ evaluate \mathcal{L}_ρ at G^\star

$$\begin{aligned} \mathcal{L}_\rho(F; \Lambda) &:= \mathcal{L}_\rho(F, G^\star(F, \Lambda); \Lambda) \\ &= J(F) + \gamma M_{(\gamma/\rho)g}(F + (1/\rho)\Lambda) - \frac{1}{2\rho} \|\Lambda\|_F^2 \end{aligned}$$

continuously differentiable

Method of multipliers

$$\mathbf{F}^{k+1} = \operatorname*{argmin}_F \mathcal{L}_{\rho^k}(\mathbf{F}; \Lambda^k)$$

$$\Lambda^{k+1} = \gamma \nabla M_{(\gamma/\rho^k)g}(\mathbf{F}^{k+1} + (1/\rho^k)\Lambda^k)$$

- FEATURES

- ★ outstanding practical performance
- ★ nonconvex J : convergence to a local minimum
- ★ F -minimization: differentiable problem
- ★ adaptive ρ -update

Dhingra & Jovanović, ACC '16

Dhingra & Jovanović, arXiv:1610.04514

- **G -UPDATE IN SPARSITY-PROMOTING PROBLEM**

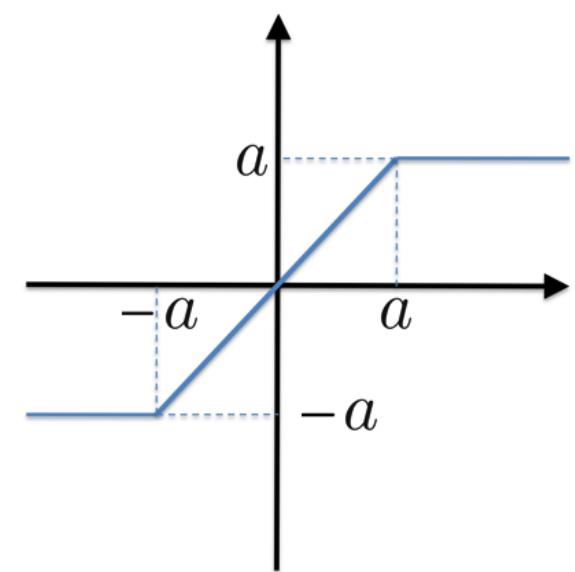
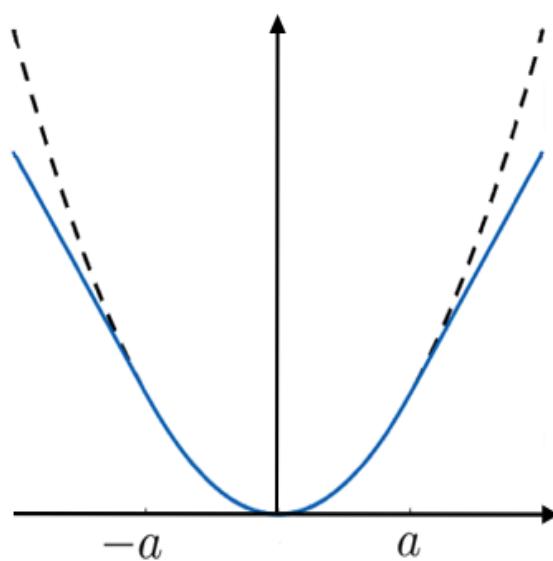
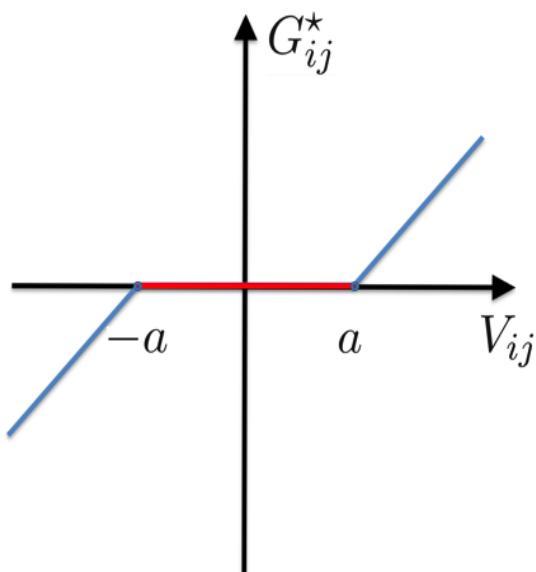
$$\underset{\substack{G_{ij} \\ i, j}}{\text{minimize}} \quad \gamma w_{ij} |G_{ij}| + \frac{\rho}{2} (G_{ij} - V_{ij})^2$$

separability \Rightarrow **element-wise analytical solution**

prox operator
soft-thresholding

Moreau envelope
Huber function

∇M
saturation



$$a = (\gamma/\rho) w_{ij}$$

Related effort

- SPARSITY-PROMOTING H_∞ CONTROL

Schuler, Li, Lam, Allgöwer, IJC '11

Schuler, Münz, Allgöwer, IFAC '12

- SYSTEMS WITH SYMMETRIES

Dhingra & Jovanović, ACC '15

Wu & Jovanović, SCL '17

- CONVEX RELAXATIONS

Lavaei, Allerton '13

Fazelnia, Madani, Lavaei, CDC '14

Fardad & Jovanović, ACC '14

- ATOMIC NORM REGULARIZATION

Matni, CDC '13; IEEE TCNS '17; Matni & Chandrasekaran, IEEE TAC '16

- SYSTEM-LEVEL SYNTHESIS

Wang, Matni, Doyle, IEEE TAC '17 (submitted)

Summary

- SPARSITY-PROMOTING OPTIMAL CONTROL

- ★ Performance vs sparsity tradeoff

Lin, Fardad, Jovanović, IEEE TAC '13

Jovanović & Dhingra, EJC '16

- ★ Software

www.umn.edu/~mihailo/software/lqrsp/

- ONGOING EFFORT

- ★ Leader selection in large dynamic networks

Lin, Fardad, Jovanović, IEEE TAC '14

- ★ Optimal synchronization of sparse oscillator networks

Fardad, Lin, Jovanović, IEEE TAC '14

- ★ Optimal design of distributed integral action

Wu, Dörfler, Jovanović, ACC '16

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UTRC



Neil
U of M



Xiaofan
Siemens



Sepideh
USC



Florian
ETH Zürich

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(Program manager: Kishan Baheti)

AFOSR Award FA9550-16-1-0009

(Program manager: Frederick Leve)